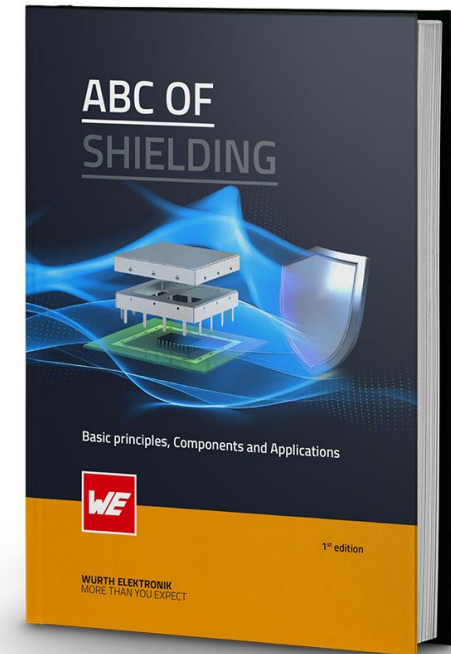
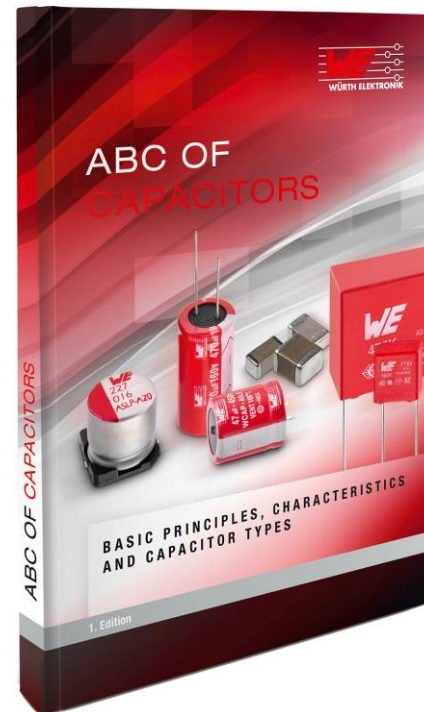




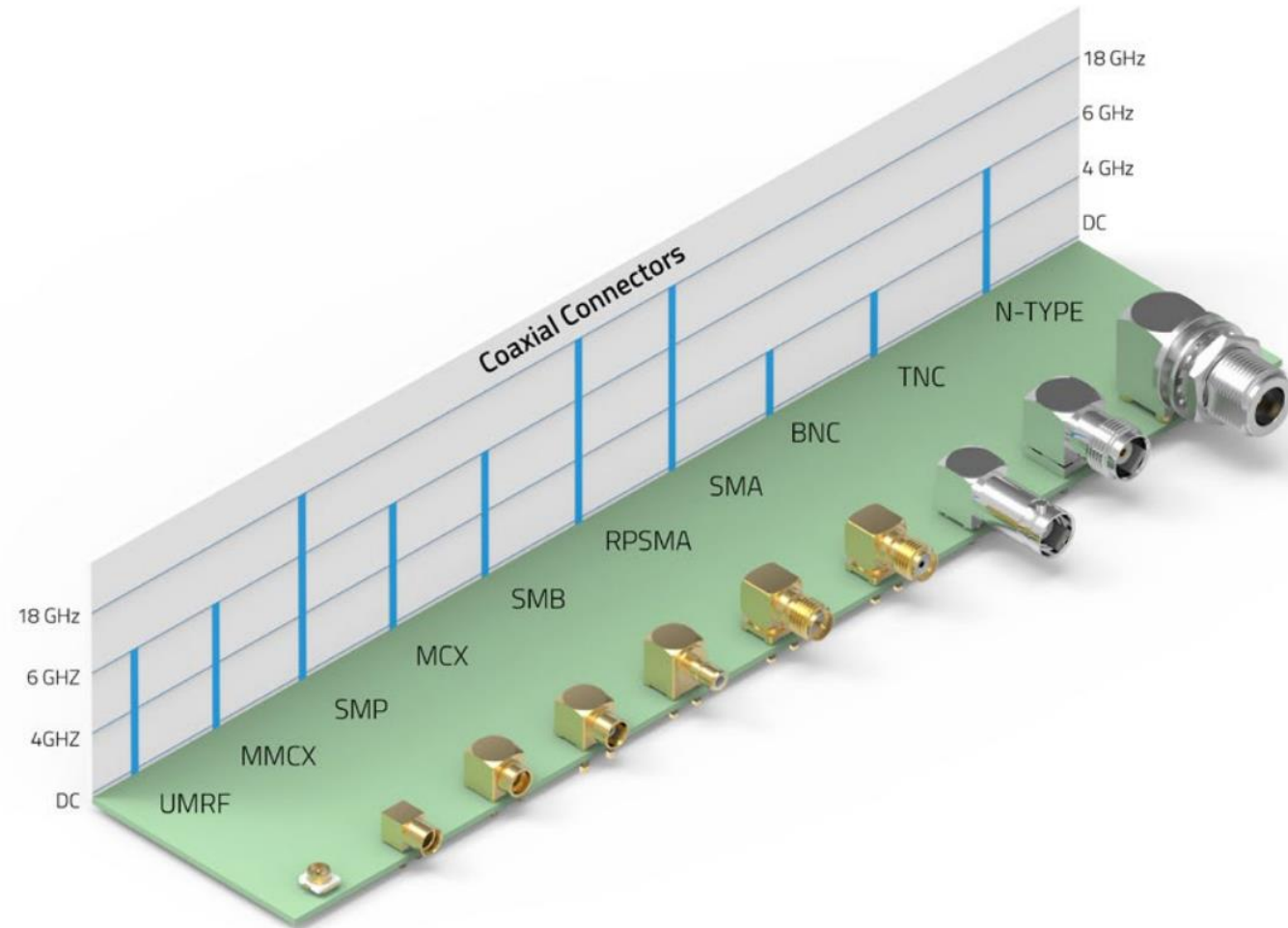
FUNCTION AND CONSTRUCTION OF COAXIAL CONNECTORS, AND THE GOLDEN RULES OF THEIR USE IN PCB DESIGN

Mateusz Stępień, FAE

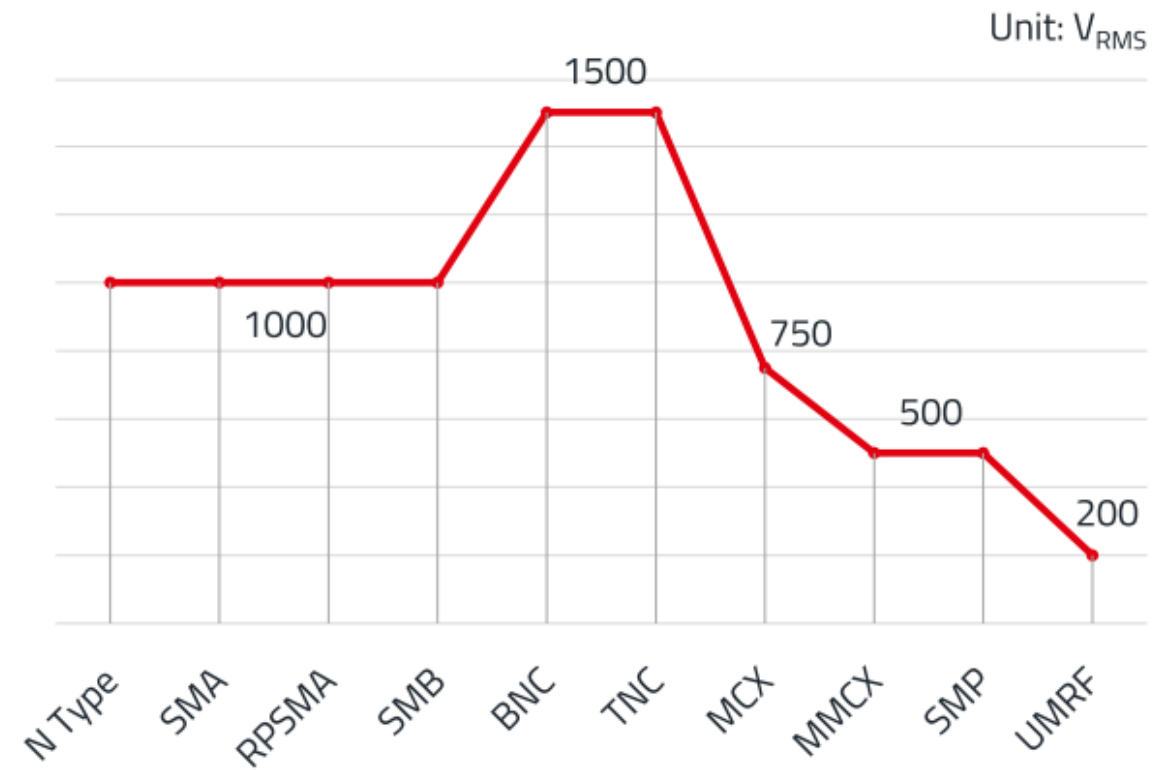
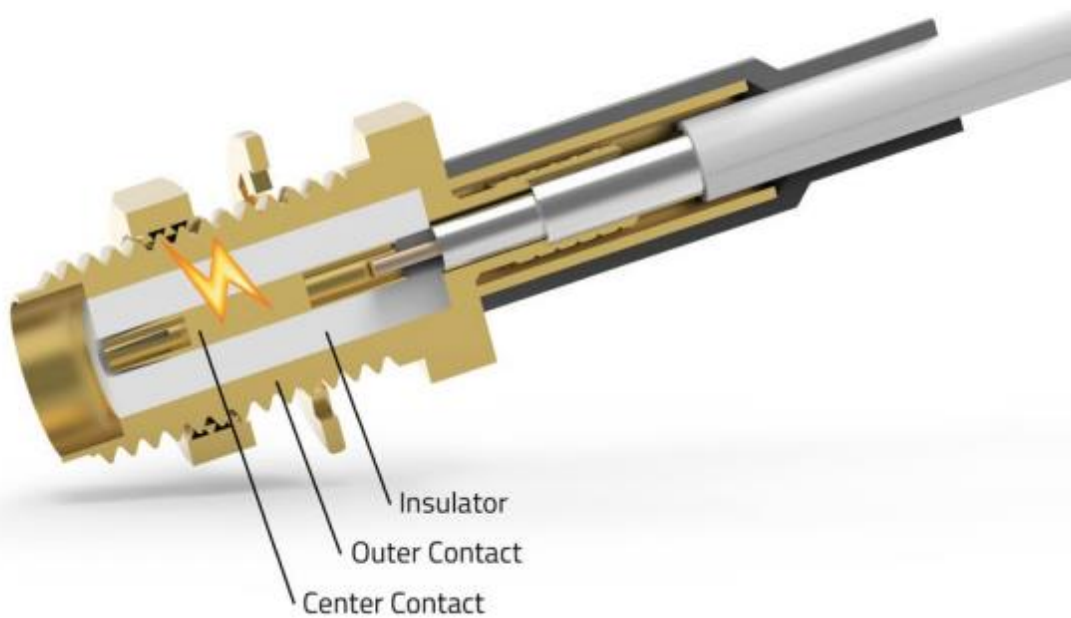




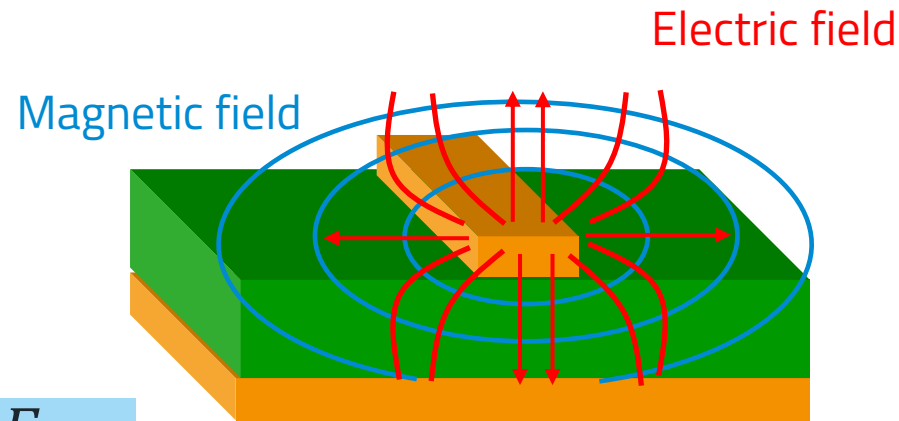
COAXIAL CONNECTORS



VOLTAGE HANDLING

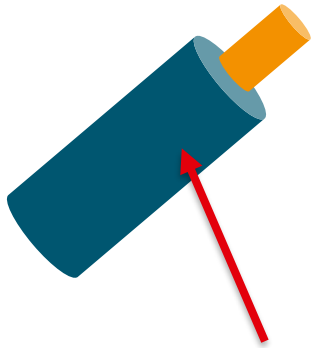


PERMITTIVITY



$$Z_c = \frac{E}{H}$$

$$Z_c = \sqrt{\frac{\mu_0 \cdot \mu_r}{\epsilon_0 \cdot \epsilon_r}}$$



ϵ_0 : vacuum electrical permittivity (**Fm⁻¹**)

ϵ_r : dielectric relative permittivity

Material	Relative permittivity ϵ_r
Vacuum	1
Air	≈ 1
PTFE (Teflon)	2.1
PET	2.3
Roger (RO4003)	3.6
FR4	4.5
PVC	5
glass	≈ 5 to 7.5
water	≈ 80 (at 20°C)

PCB CONNECTORS

- Fixed PCB thickness
- Flat tab & round post pin



End launch



Through hole

- Good retention on PCB
- Wider range of PCB thickness
- Available with SMT & THR signal pin

- Full SMT
- PCB edge card

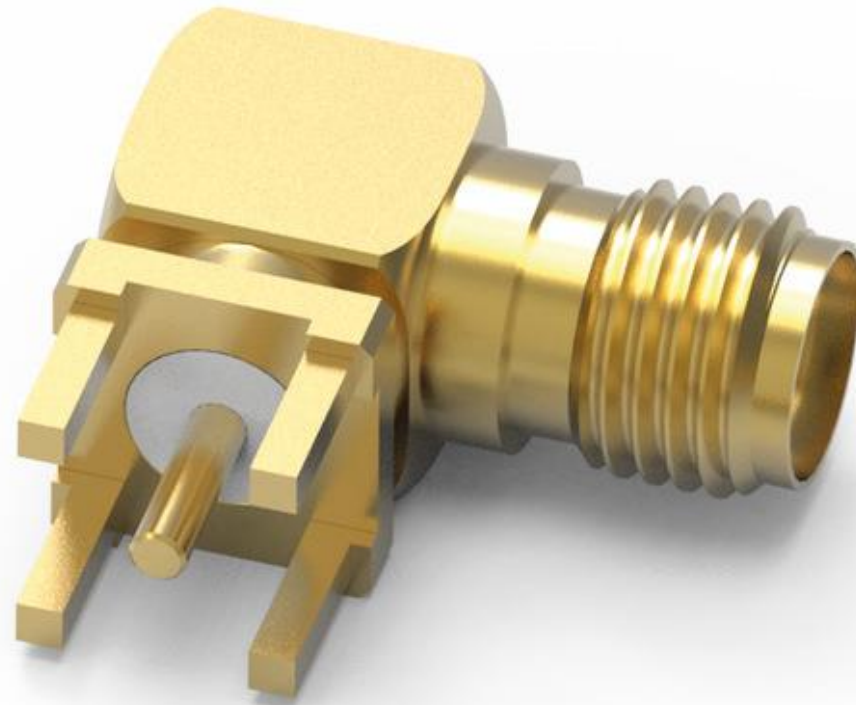


Surface mount



Surface mount

- Full SMT



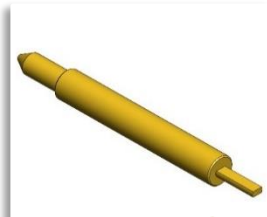
PCB CONNECTORS

Center contact back shape



Round post

- Good contact
- Low frequency



Flat tab

- Easy PCB design
- High frequency

PCB thickness



1.1mm

&

1.6mm

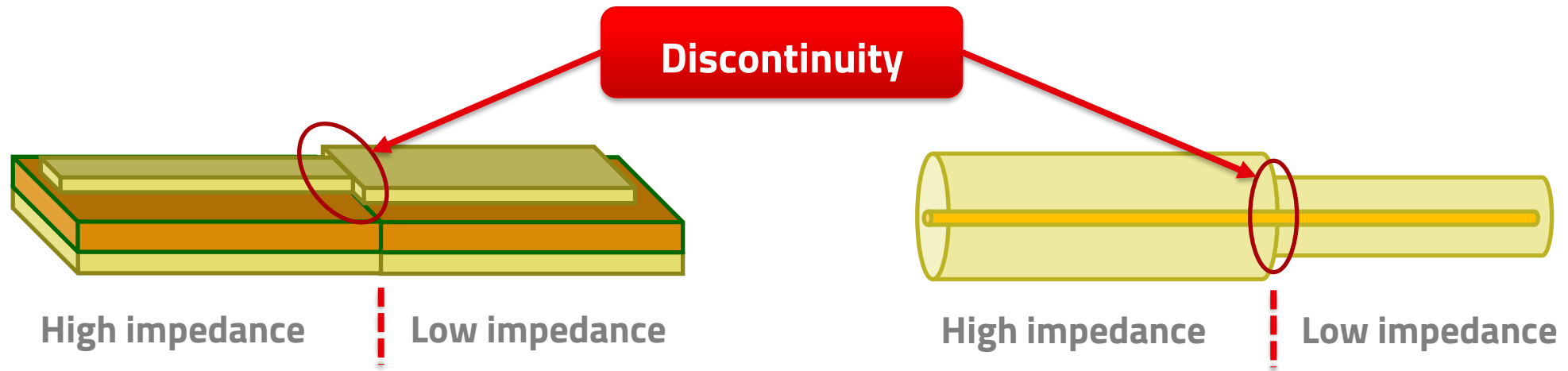


Up to 2.8mm



DISCONTINUITY IMPACT

➤ Impedance variation on RF-path

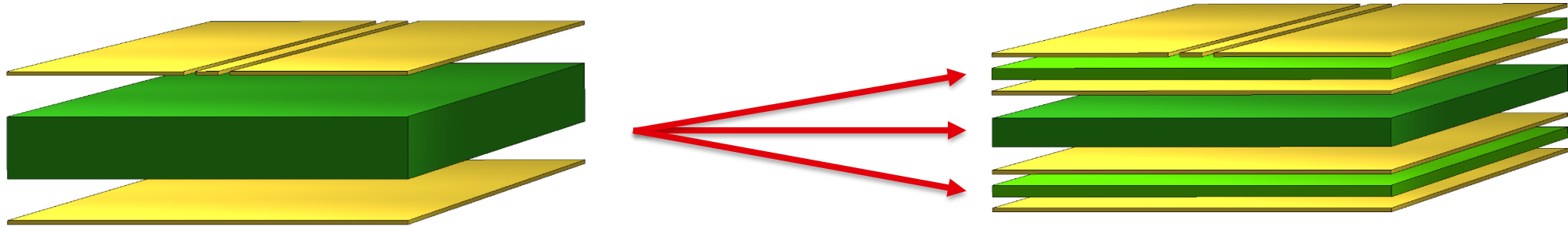


➤ Causes:

- Reflection
- Field conversion



TYPICAL LAYER SETUP



2-Layer	
Height	1.55 mm
Prepreg + core	FR4

4-Layer	
Height	1.55 mm
Prepreg + core	FR4

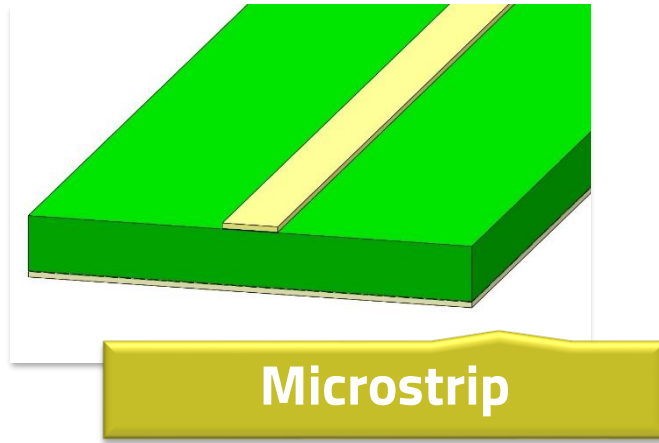
➤ Low production cost

- Mostly no separate GND-Plane

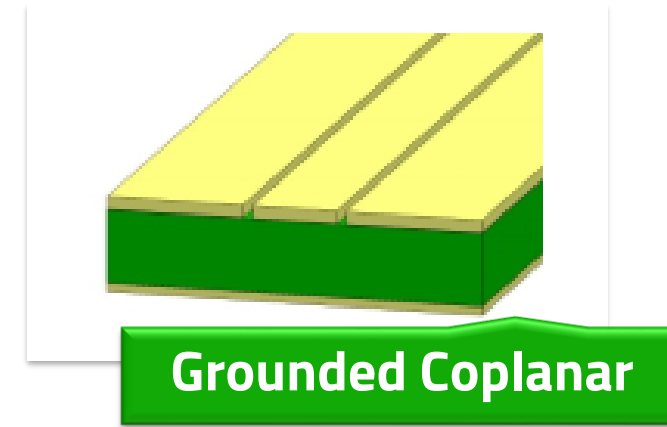
➤ Good for RF-Designs

- Separate GND-Plane
- Dielectric differs

PLANAR TRANSMISSION LINES



- Wider line width
- Antenna fed-line
- No Ground on RF-Layer
- Line width depends on substrate height and ϵ_r

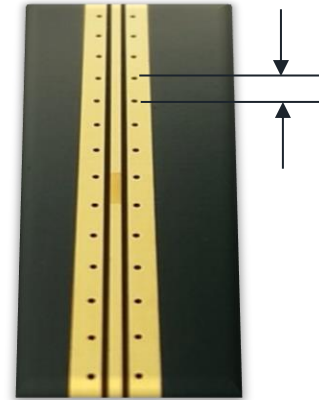
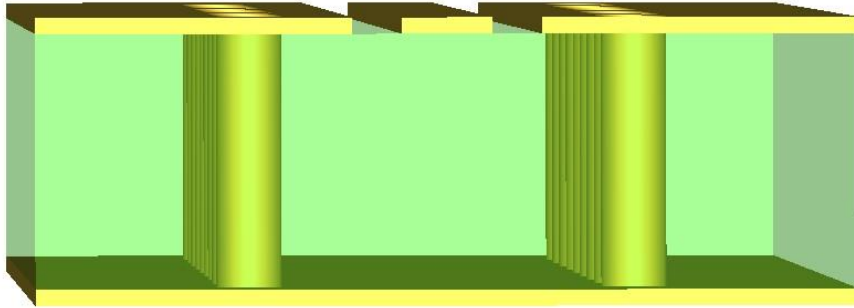


- Smaller line width
- Ground connection to components
- Various planar matching designs
- Line width depends on substrate height, ϵ_r and gap width

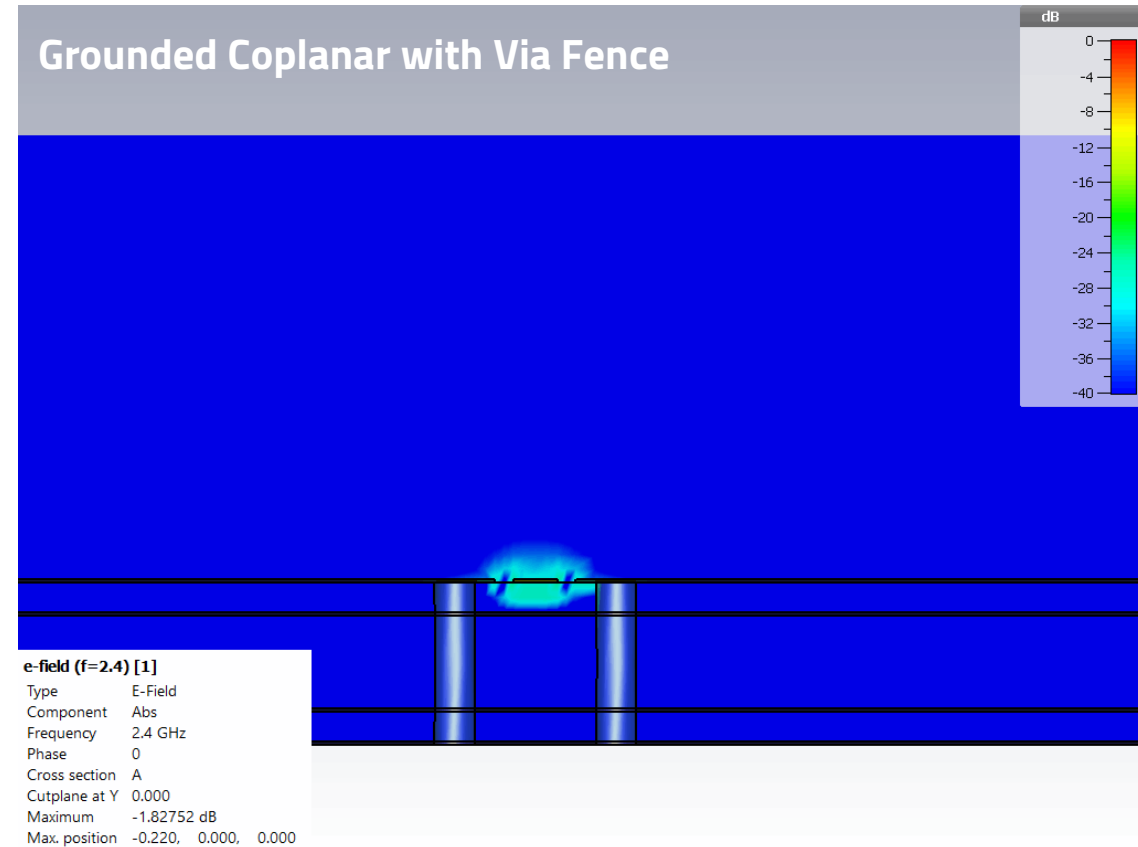
OVERVIEW PCB STRUCTURES: TIPS & TRICKS

VIA FENCE

Via center distance $\sim \lambda/10$

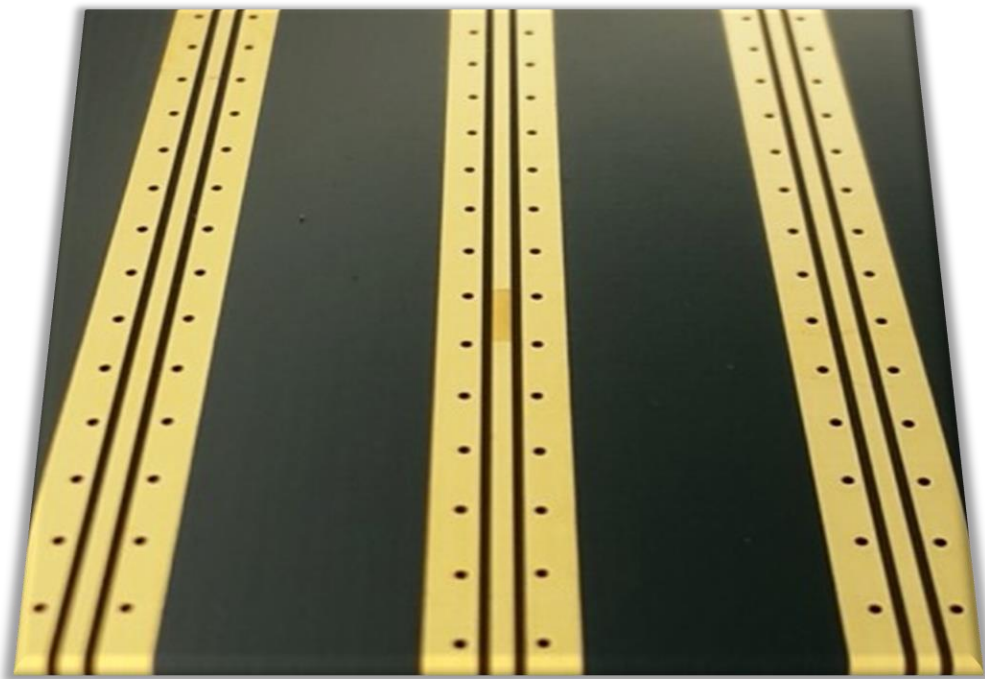


- **Field captured between GND**
 - Reduces coupling
 - Less loss
- **Stabilized ground planes**



OVERVIEW PCB STRUCTURES: TIPS & TRICKS

SOLDER-RESIST FREE



➤ Solder resist:

- Increases loss
- Adds dielectric

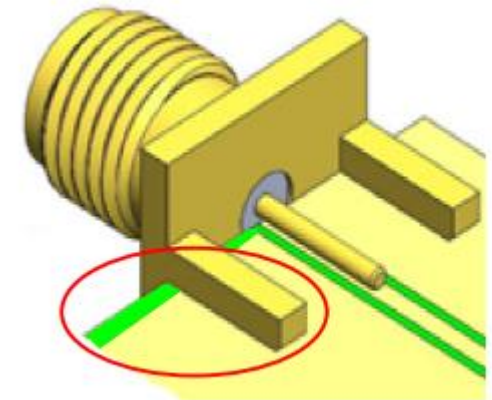
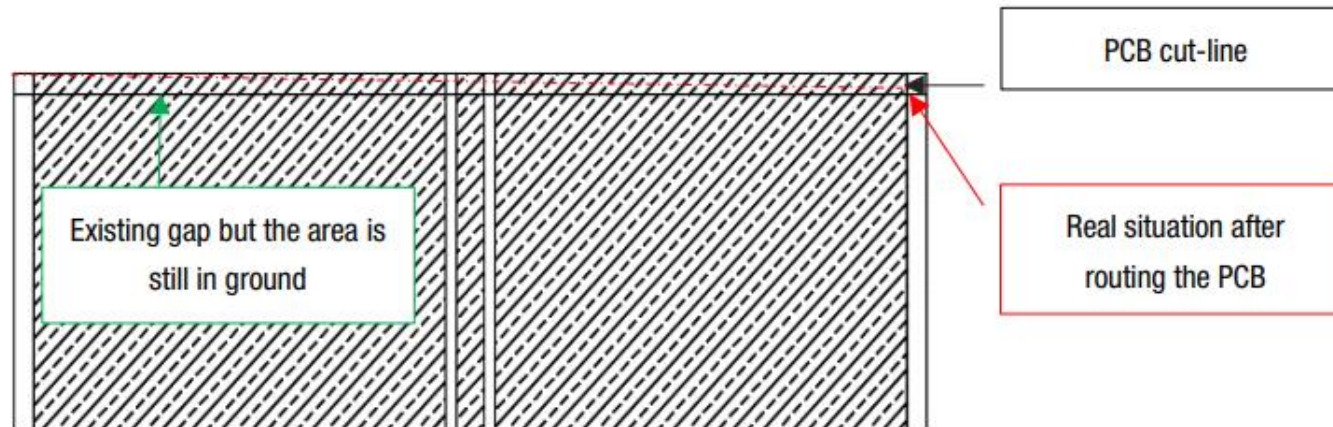
➤ Remove solder resist from:

- RF-Line and
- Near Ground-Plane

COMBINATION CONNECTOR & PCB: EXAMPLES

CPWG - ROUND: DESIGN - DGS

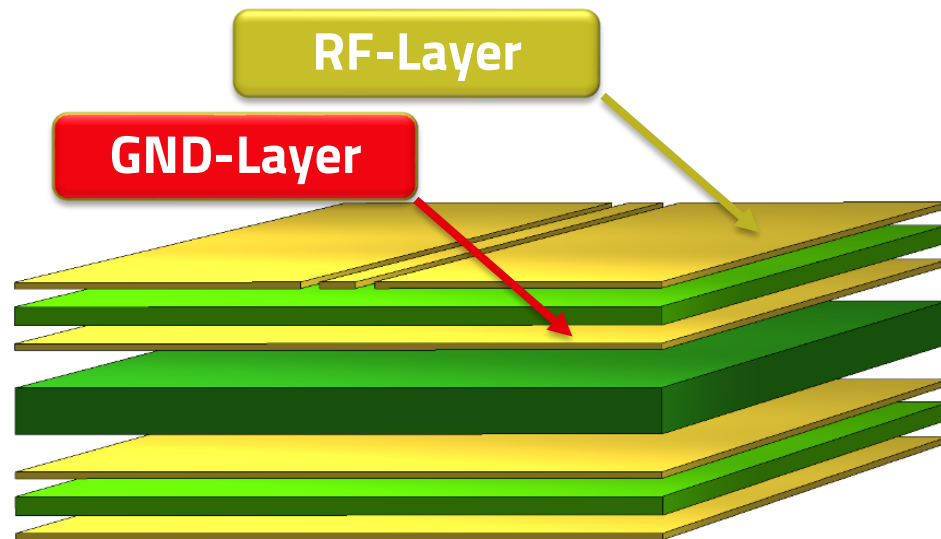
➤ PCB edge



After soldering the connector, a gap exist without ground

OVERVIEW PCB STRUCTURES: TIPS & TRICKS

LAYERS

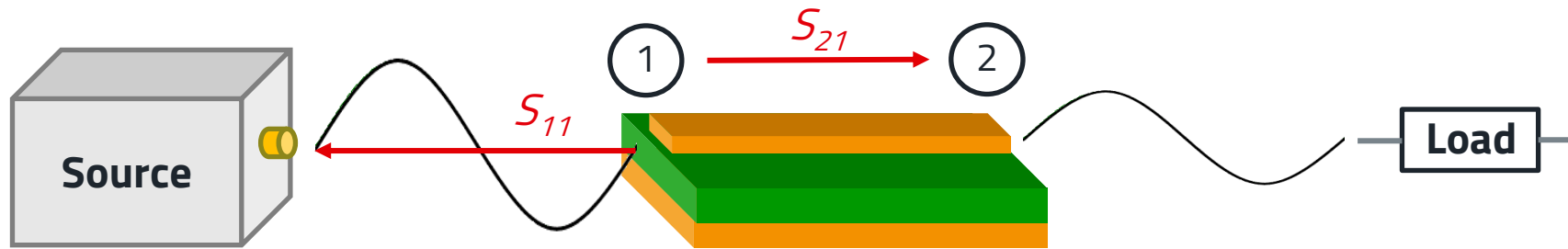


- **Top or Bottom as RF-Layer**
 - Reduced loss
 - No Vias inside trace → Mismatch!
- **Seperate ground-layer underneath RF**
 - Decreases discontinuities
 - Good connection to RF-Layer GND needed!

S PARAMETERS

$$S_{ji} = 10 \log \frac{\text{power } j}{\text{power } i} \text{ (dB)}$$

➤ Factor of reflection and throughput:



➤ S-Matrix:

- S_{ii} : Reflection at Port i
- S_{ji} : Insertion losses from Port i to Port j

➤ Power domain!

➤ Mostly given in dB:

- e.g. $-3\text{dB} = 10 \times \log(0.5)$

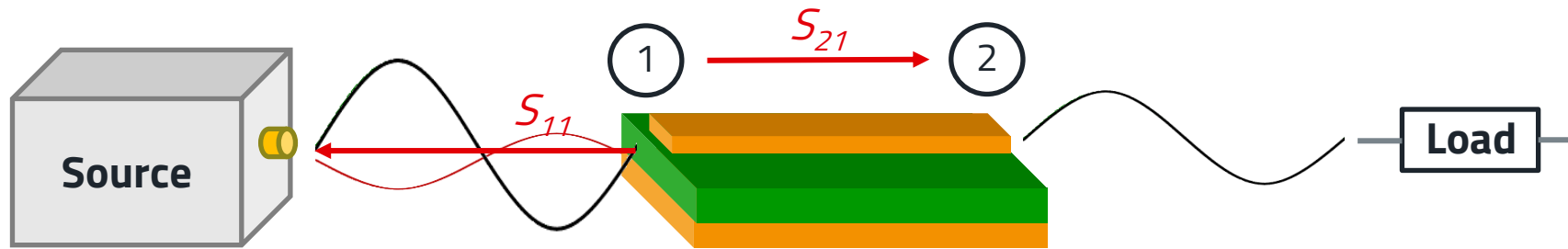
Signal power-50%

➤ Calculation from source:

- $1000 \text{ mW} \times 0.5 = 500\text{mW}$
- $30\text{dBm} - 3\text{dB} = 27\text{dBm}$

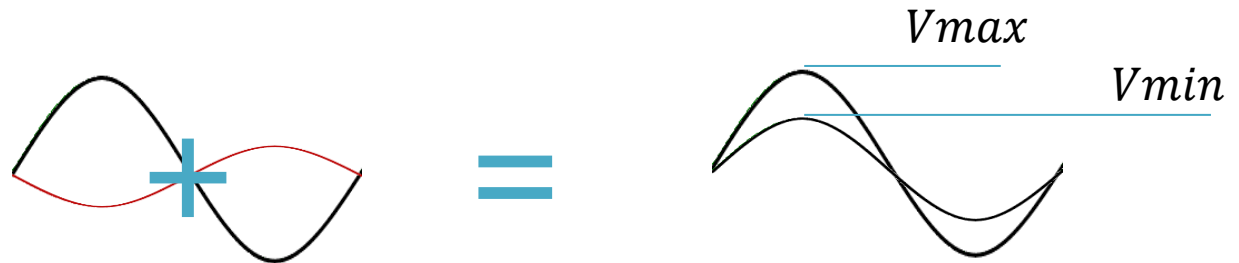
S PARAMETERS: VSWR

➤ Voltage standing wave ratio (VSWR):



➤ Only reflection/return loss:

$$VSWR = \frac{V_{max}}{V_{min}} [dB]$$



➤ Voltage domain!

S PARAMETERS & VSWR

VSWR	Return Loss (dB)	Reflection (%)	
1.0	∞	0	Very well matched
1.02	40.1	0.01	
1.03	35.3	0.03	
1.1	26.4	0.23	Well matched
1.12	24.9	0.32	
1.15	23.1	0.49	
1.2	20.8	0.83	Matched
1.3	17.7	1.70	
1.4	15.6	2.75	
1.5	14.0	3.98	Poorly matched
1.7	11.7	6.76	
1.8	10.9	8.13	
2.0	9.5	11.22	Not matched

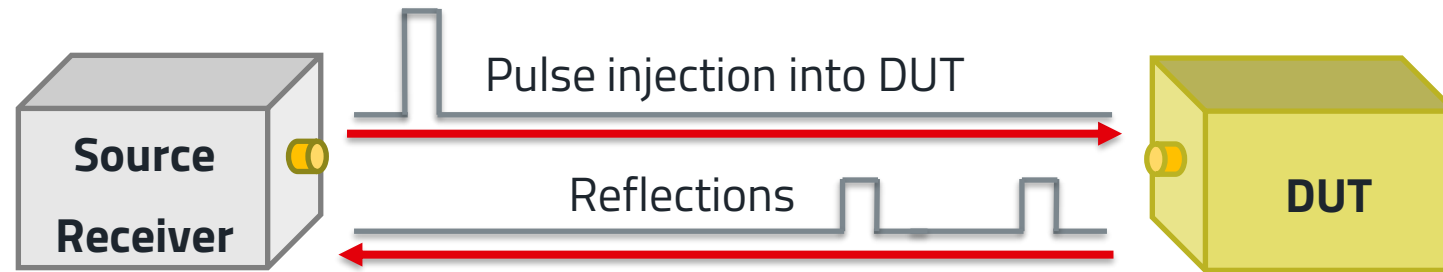
$$S_{11} = 10 \log \frac{\text{power sent 1}}{\text{power received 1}} \text{ (dB)}$$

➤ Return loss example:

- 20 dB: 99 % insertion & 1% reflection
- 15 dB: 97% insertion & 3% reflection
- 10 dB: 90% insertion & 10% reflection
- 6 dB: 75% insertion & 25% reflection

TIME DOMAIN REFLECTOMETRY

➤ Wave impedance measurement through a system



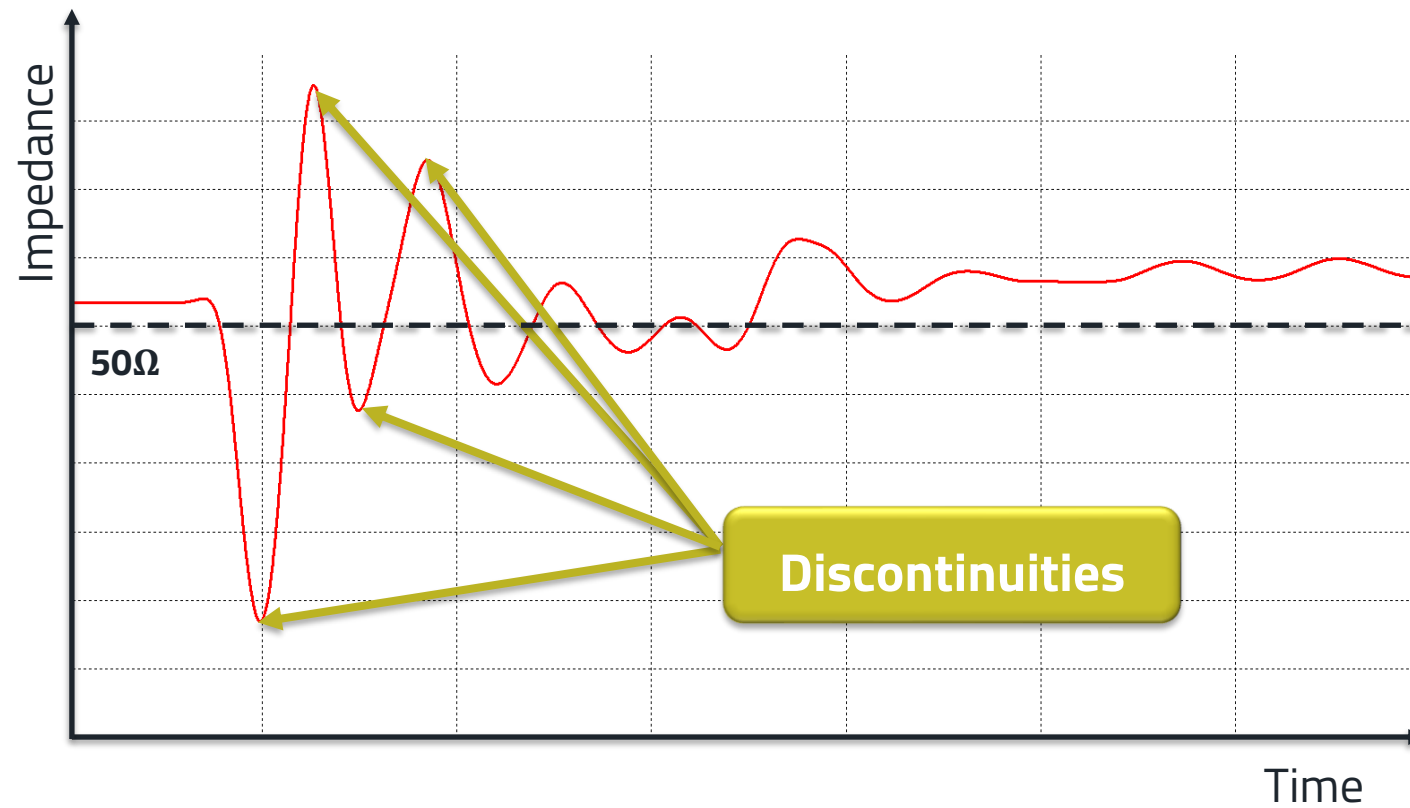
➤ Measurement of reflections

- Amplitude
- Time

VNA calculation

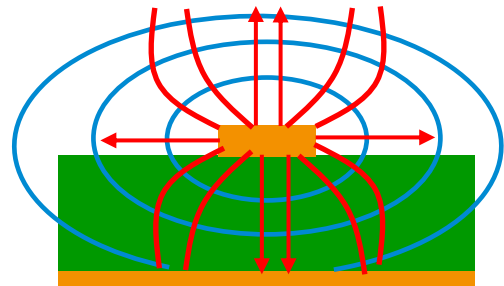
➤ Characteristic impedance

TIME DOMAIN REFLECTOMETRY

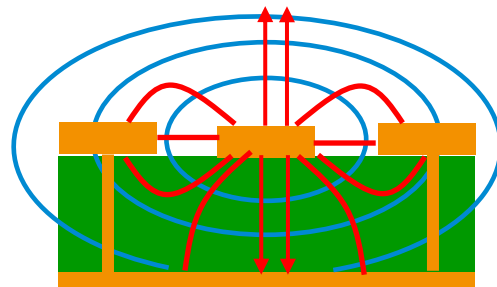


➤ Time ~ Length :
$$L = \frac{c}{\sqrt{\epsilon_r \mu_r}} \times t$$

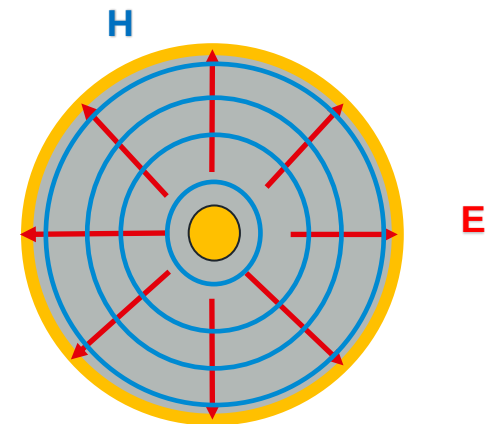
RADIATION - WAVE GUIDE LOSSES



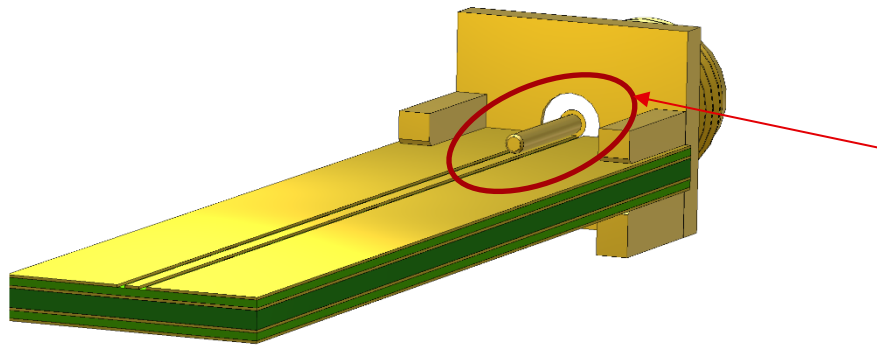
Microstrip : middle losses



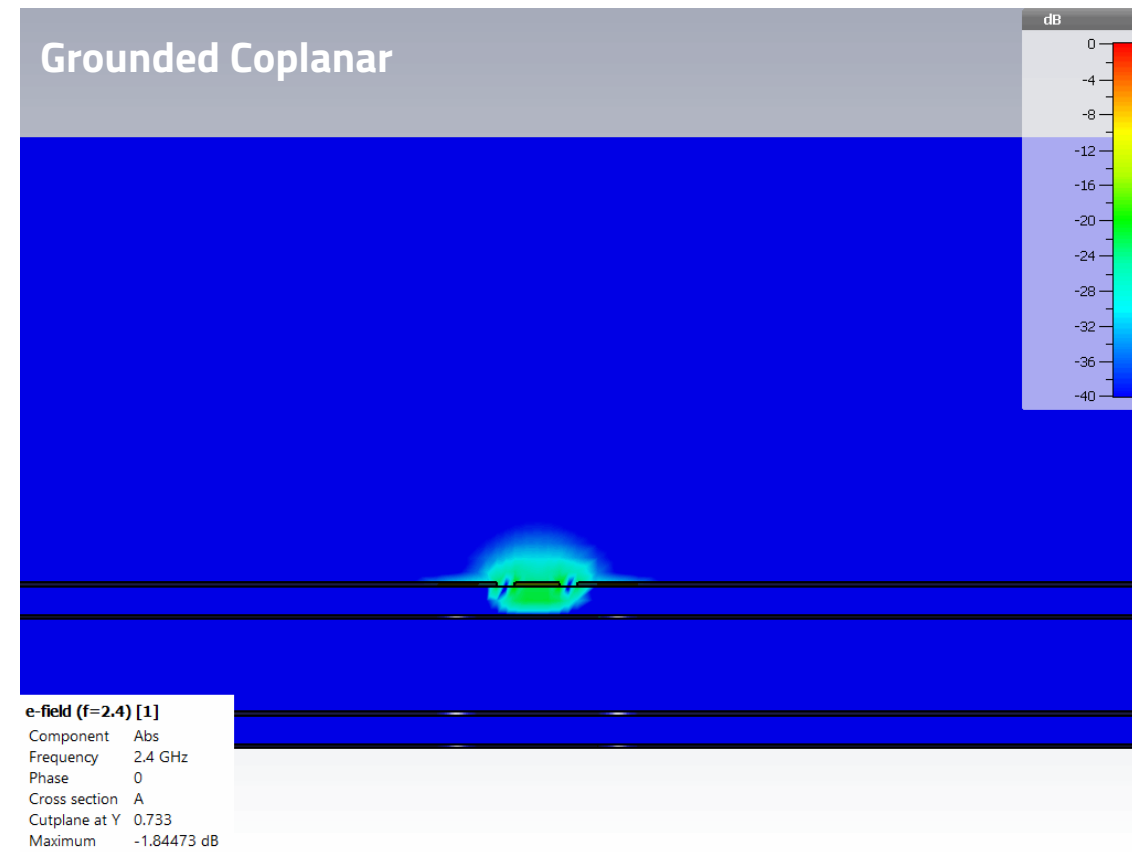
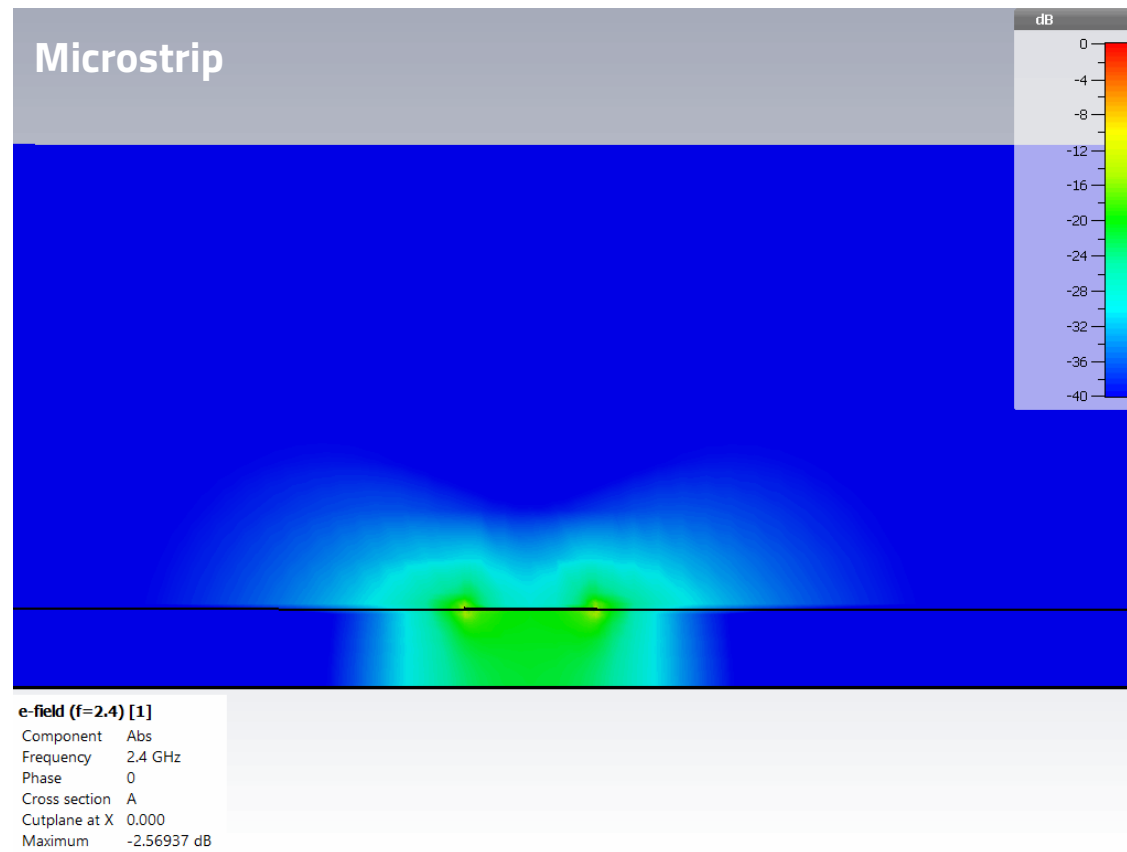
Ground coplanar: low losses



Coax: very low losses



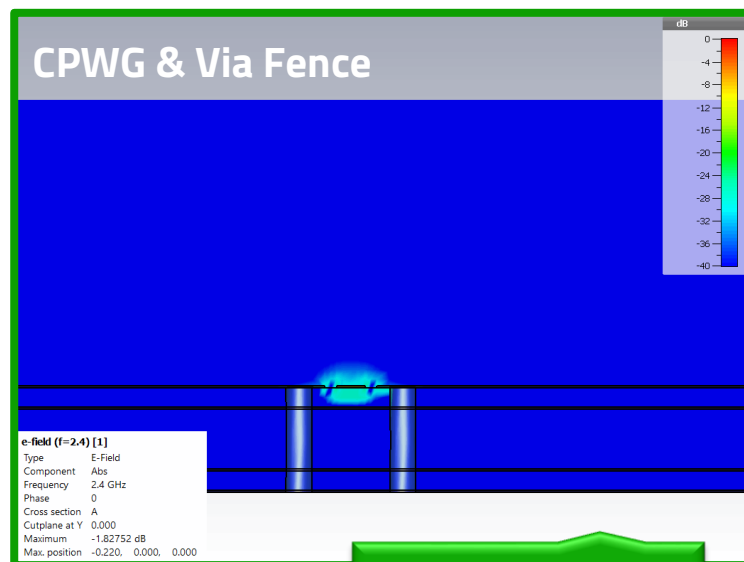
FIELD DISTRIBUTION



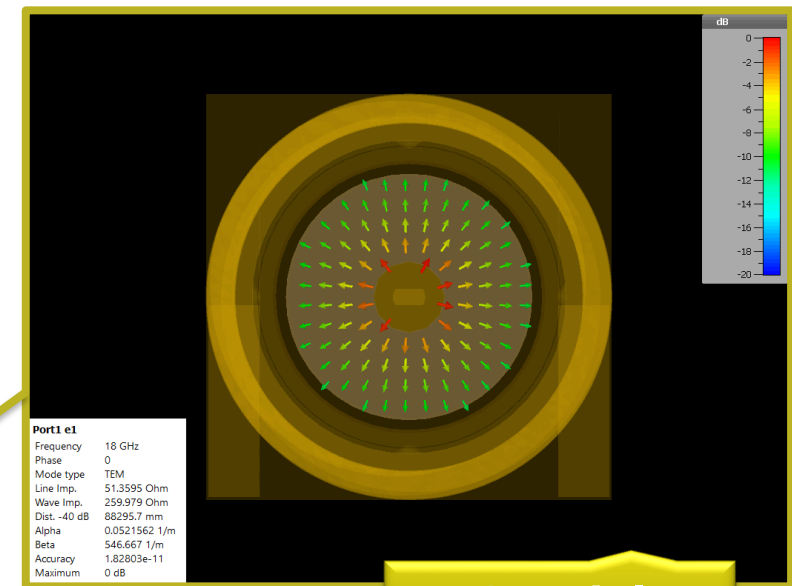
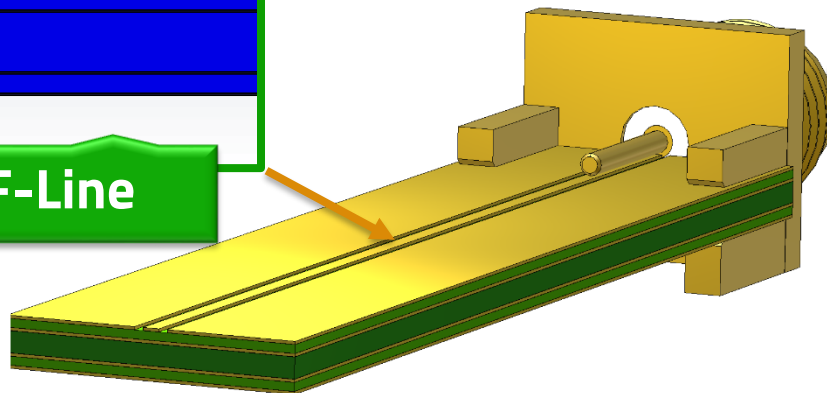
COMBINATION CONNECTOR & PCB: RF-EFFECTS

FIELD CONVERSION (1)

➤ Geometry of field changes at connection point



RF-Line



Coaxial

COMBINATION CONNECTOR & PCB: RF-EFFECTS

DISCONTINUITIES

➤ **Discontinuity at connection point**

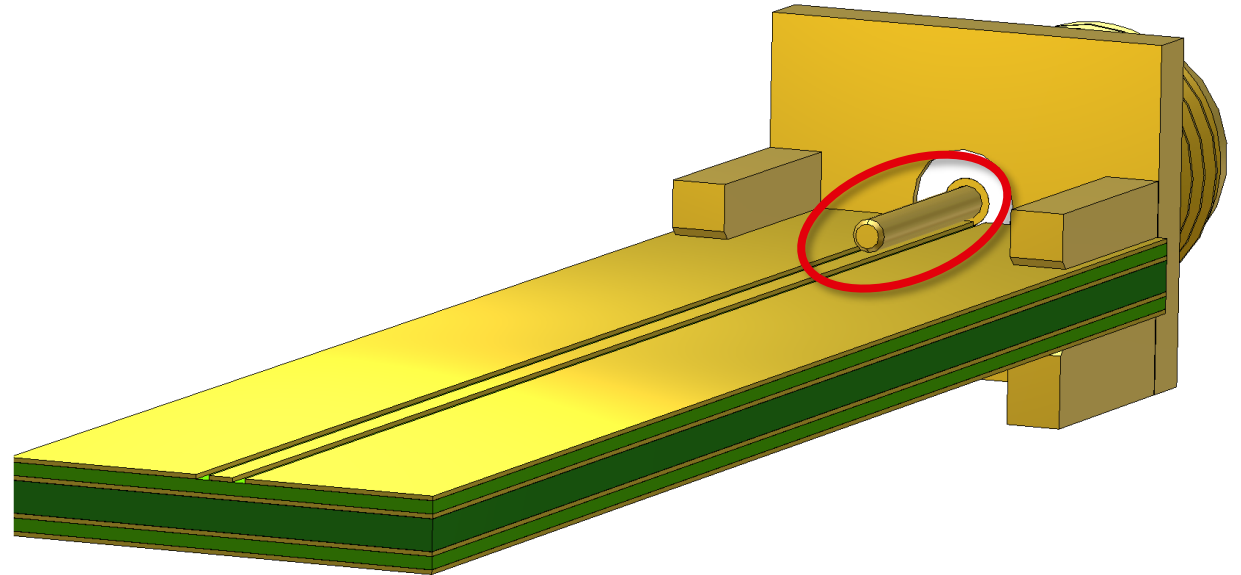
- Change in line impedance
- Field conversion



➤ **Huge Mismatch and Reflection!**



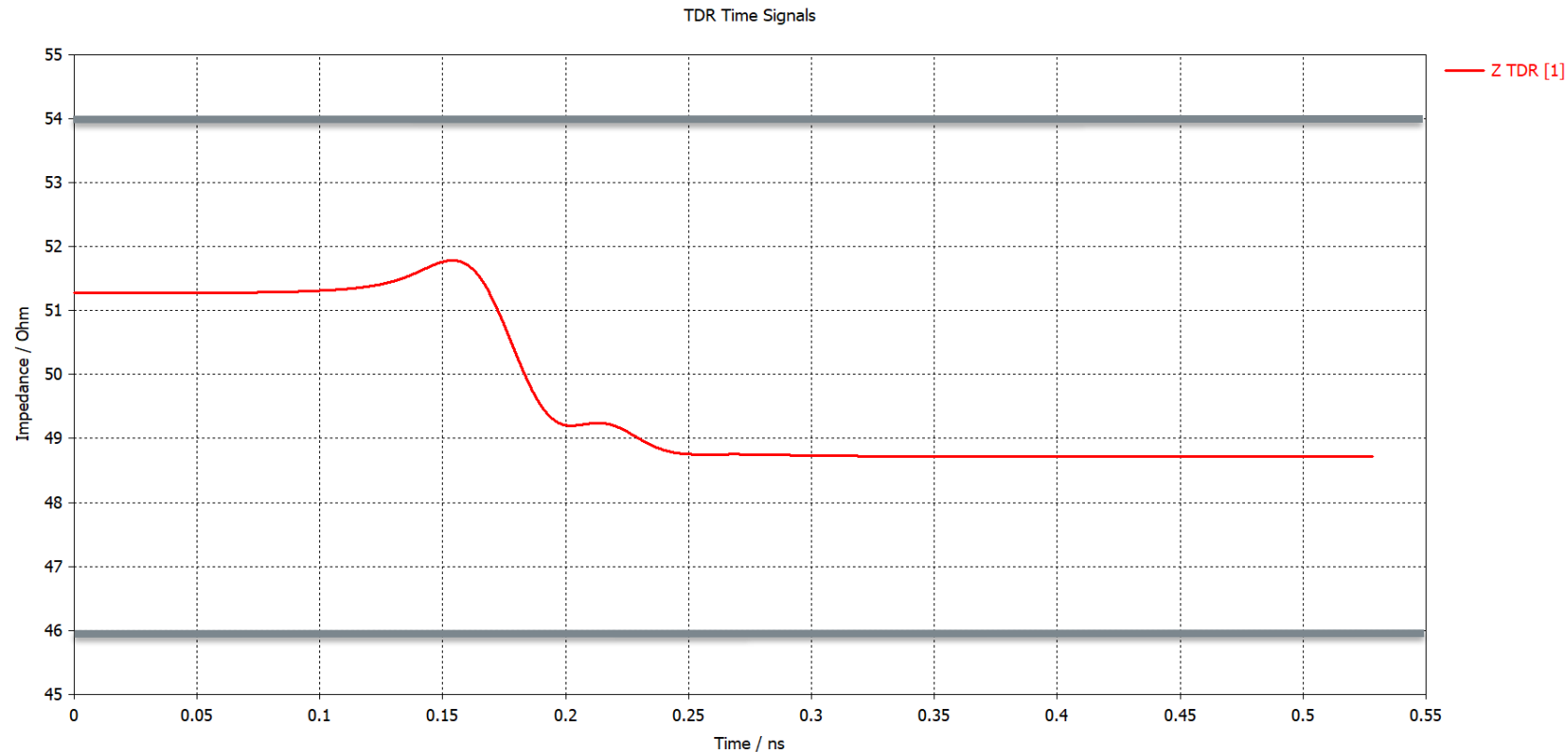
➤ **Planar matching circuit needed**



COMBINATION CONNECTOR & PCB: EXAMPLES

PROPERTIES

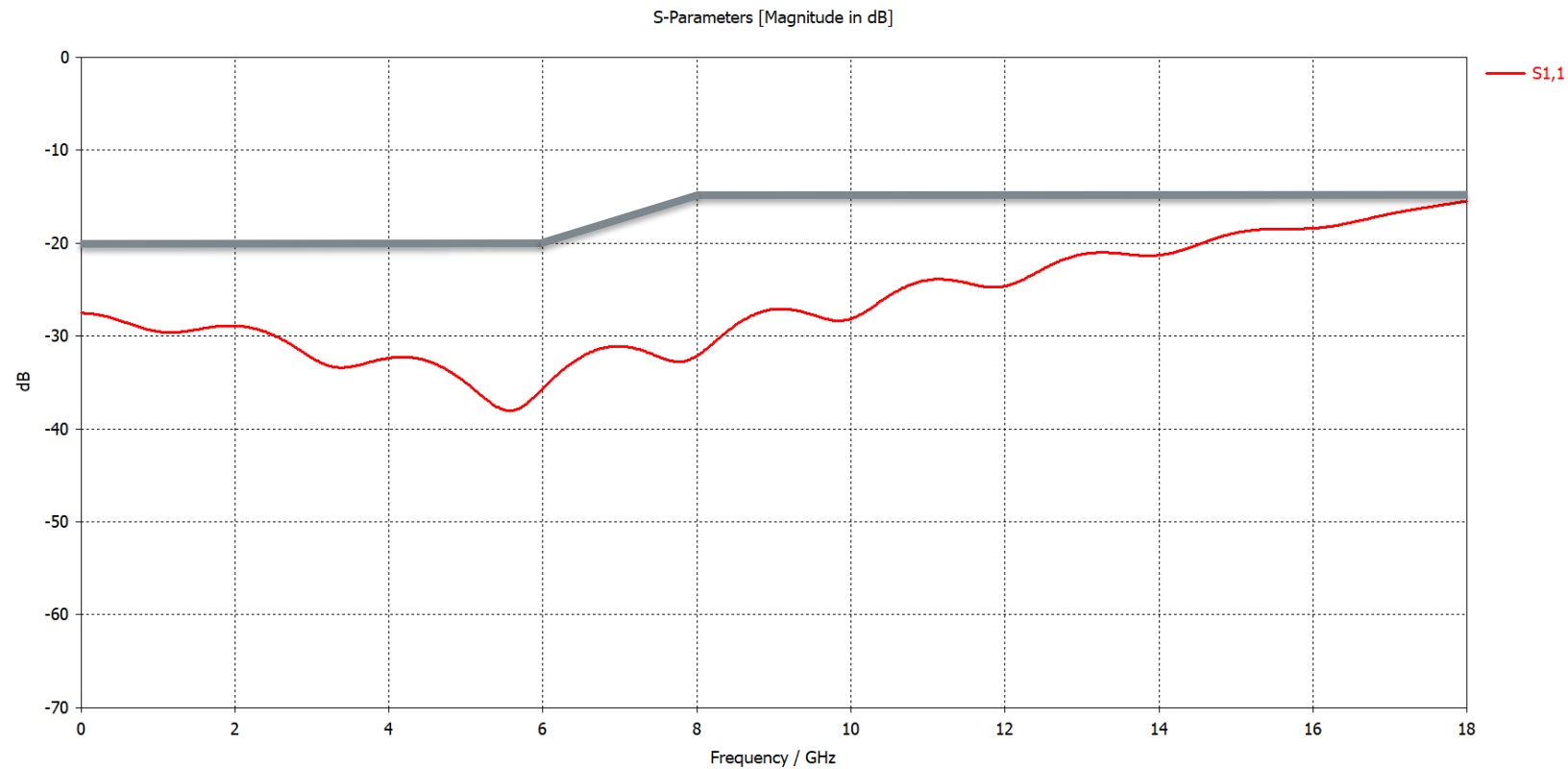
Characteristic impedance limits



COMBINATION CONNECTOR & PCB: EXAMPLES

PROPERTIES

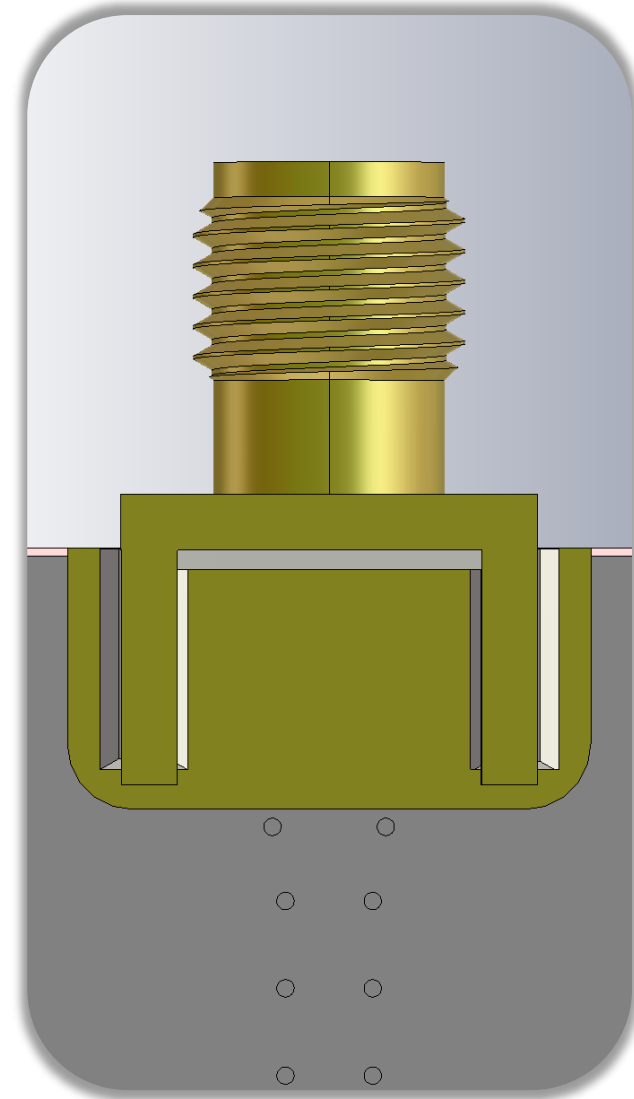
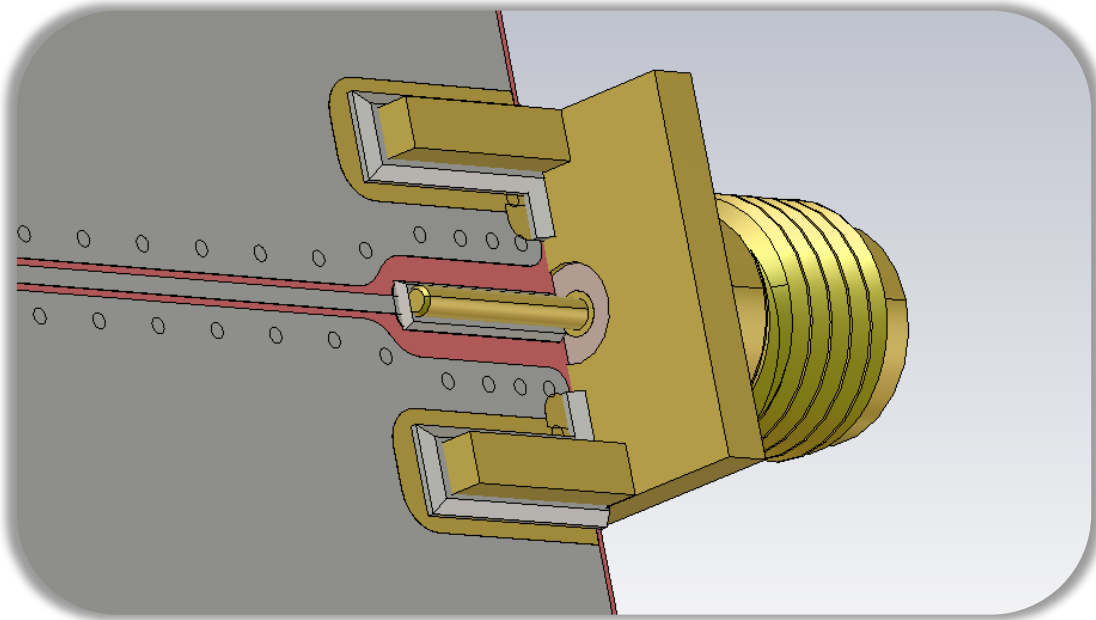
Return loss S11 limits



COMBINATION CONNECTOR & PCB: EXAMPLES

CPWG - ROUND: DESIGN

- Solder pads & several vias
- Very good GND connection
- Optimized RF-Line with tapers



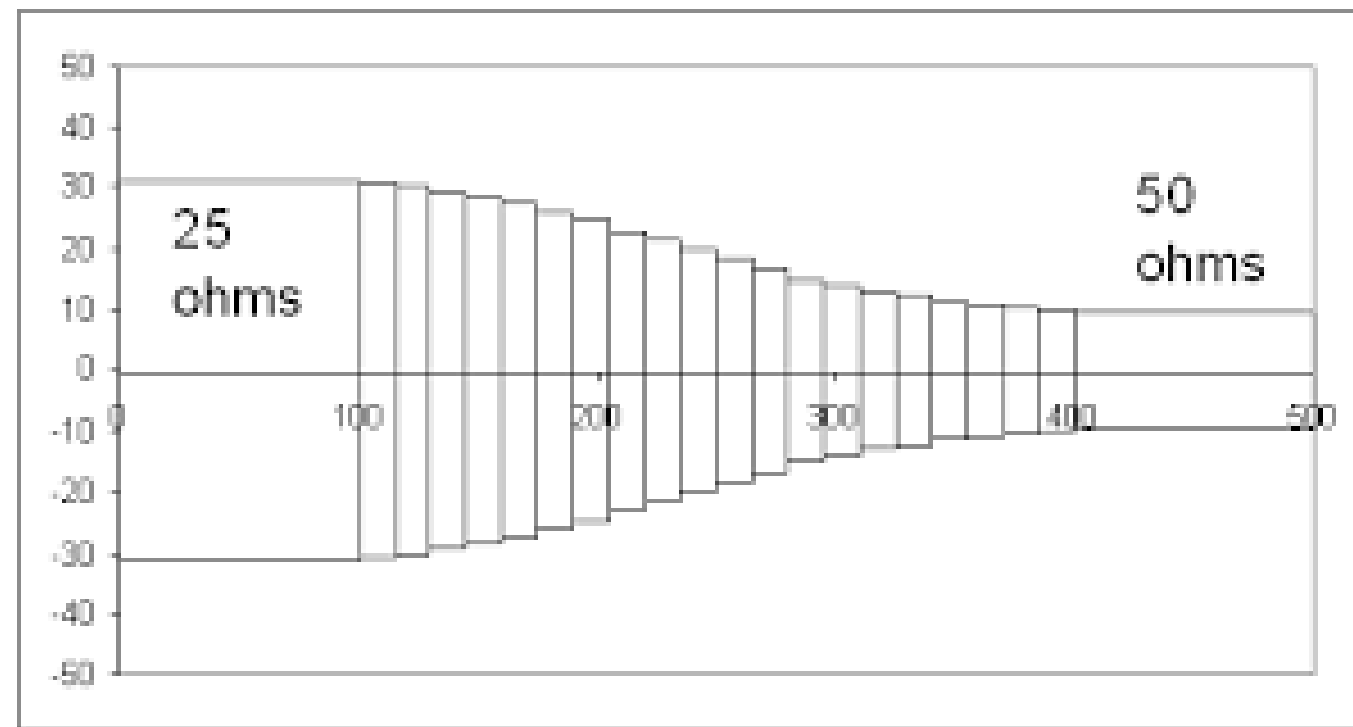
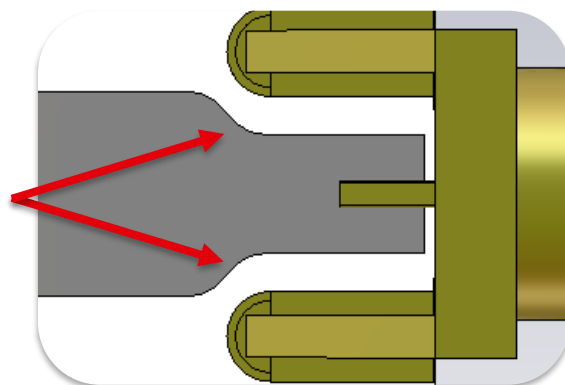
OVERVIEW PCB STRUCTURES: TIPS & TRICKS

PLANAR IMPEDANCE MATCHING

Taper Structures

Background information

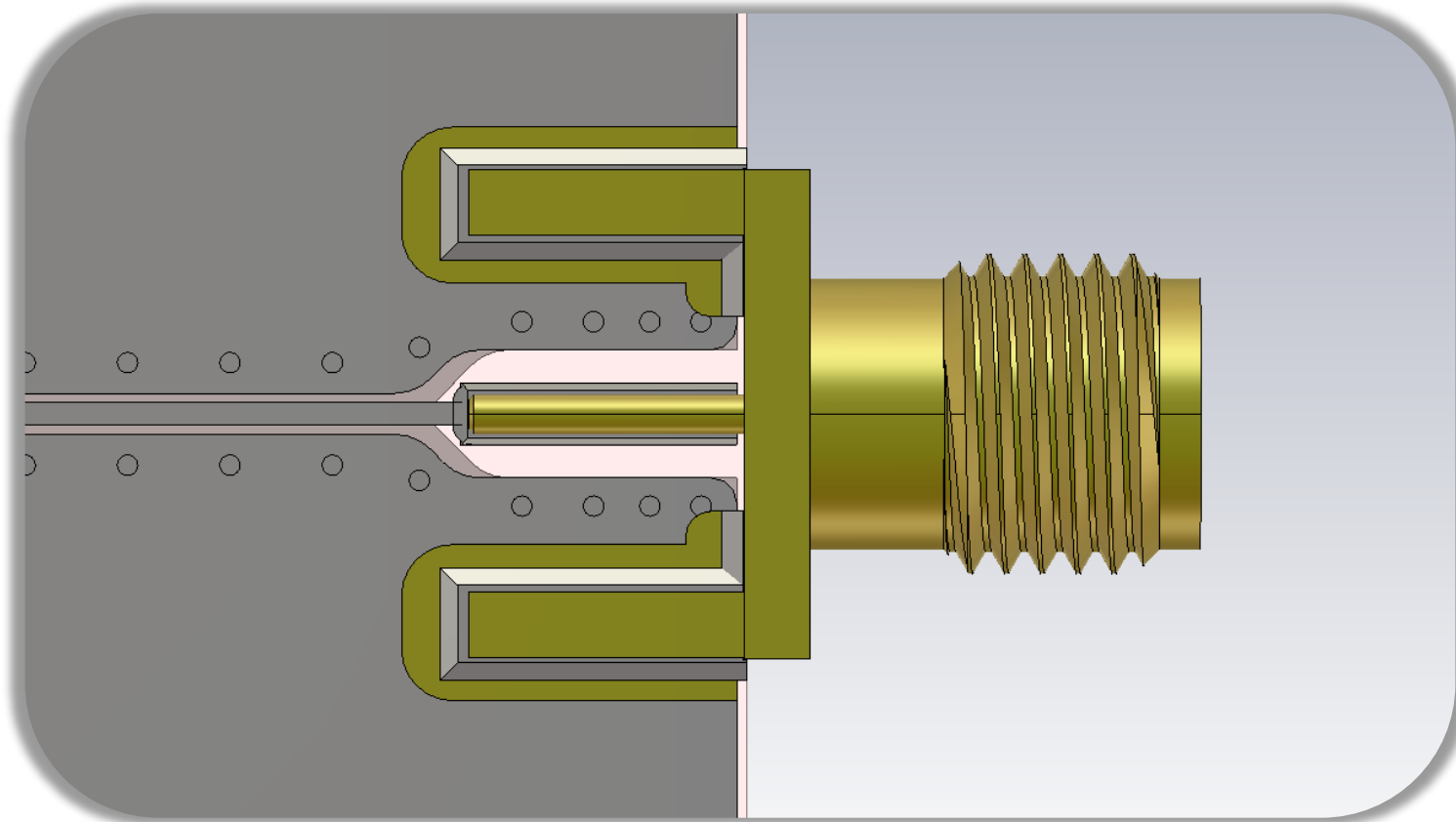
Line Taper



COMBINATION CONNECTOR & PCB: EXAMPLES

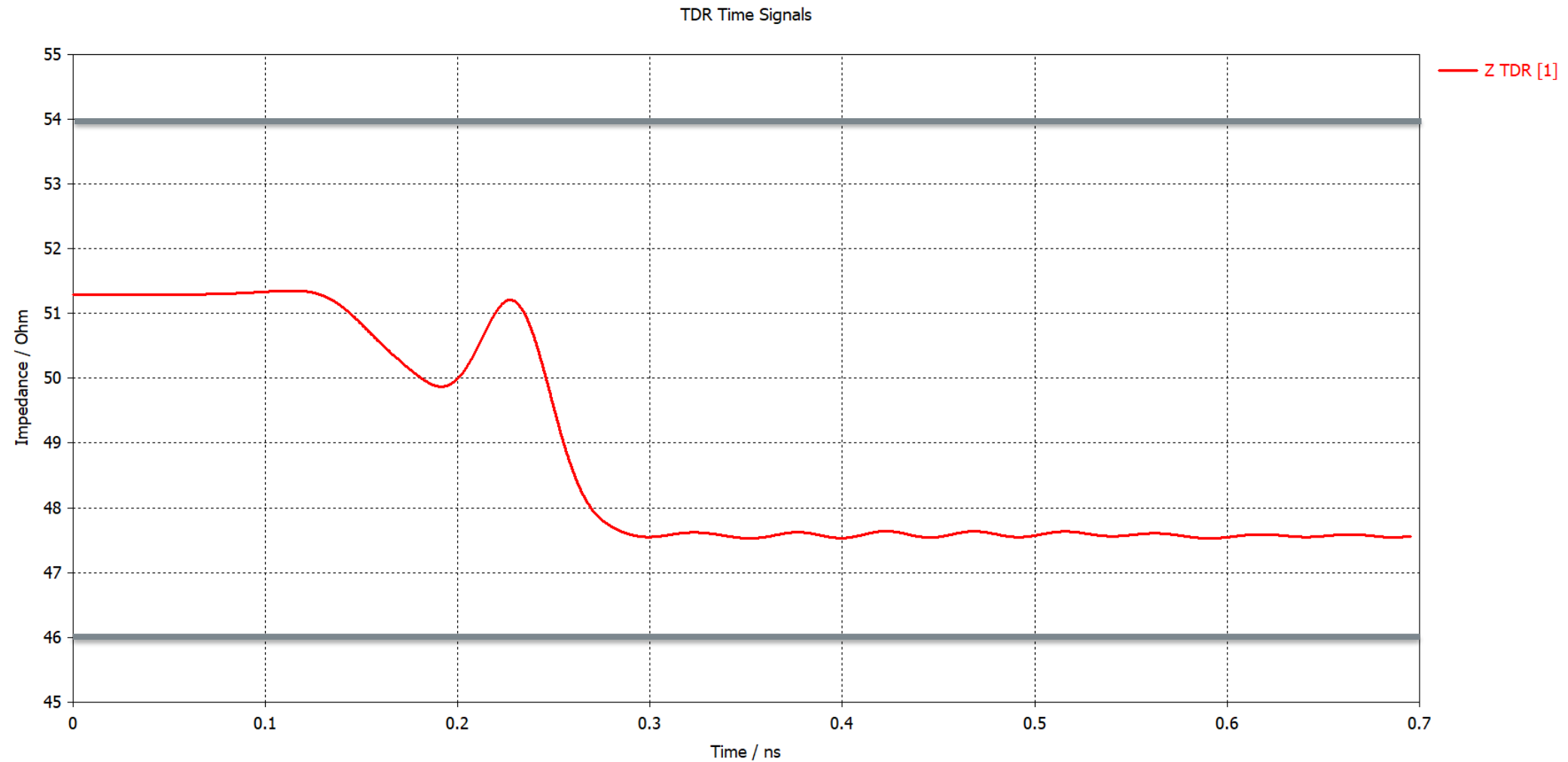
CPWG - ROUND: DESIGN - DGS

- Defective Ground Structure: matching structure → decreases parasitic capacitance



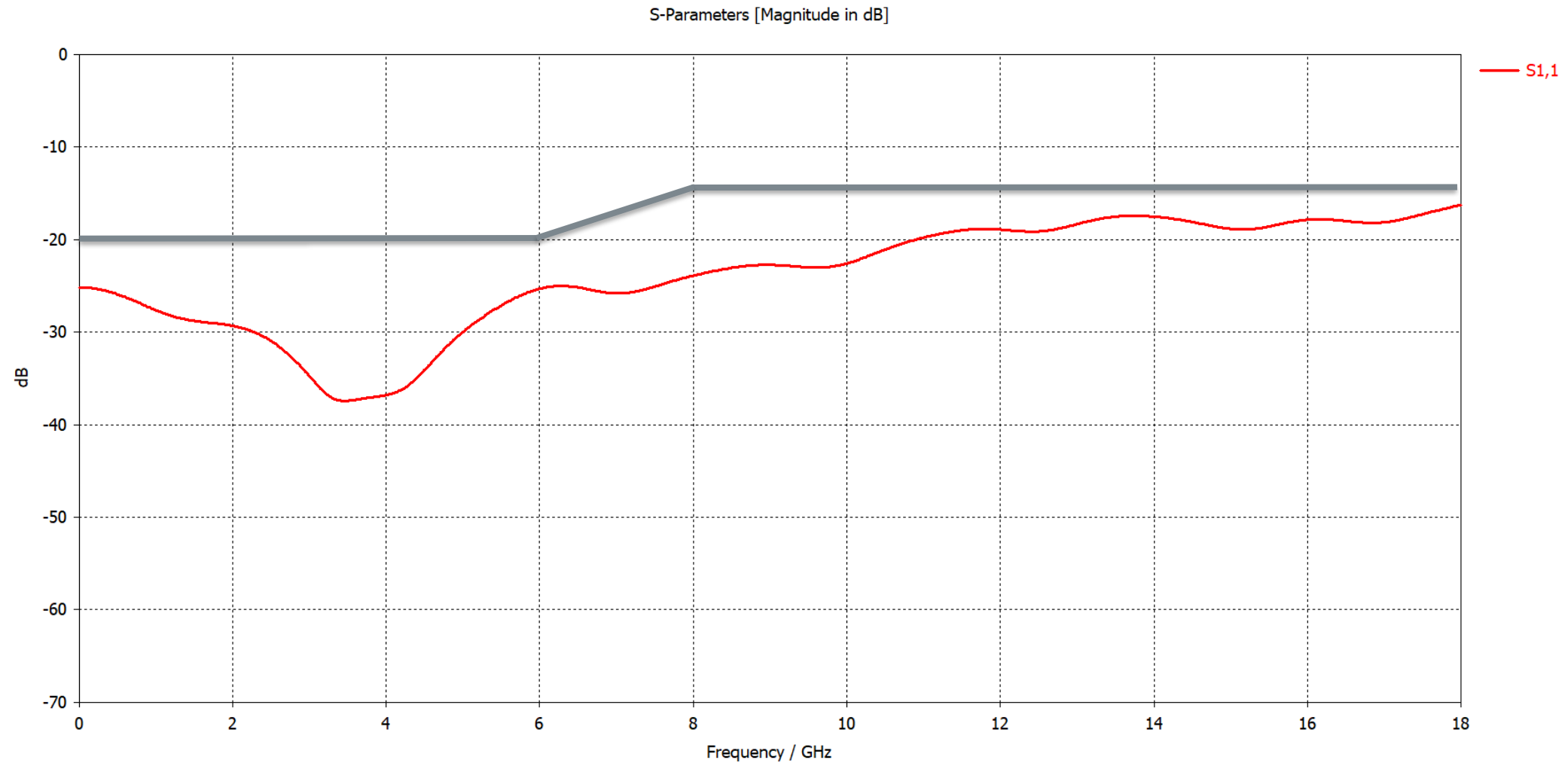
COMBINATION CONNECTOR & PCB: EXAMPLES

CPWG - ROUND: TDR



COMBINATION CONNECTOR & PCB: EXAMPLES

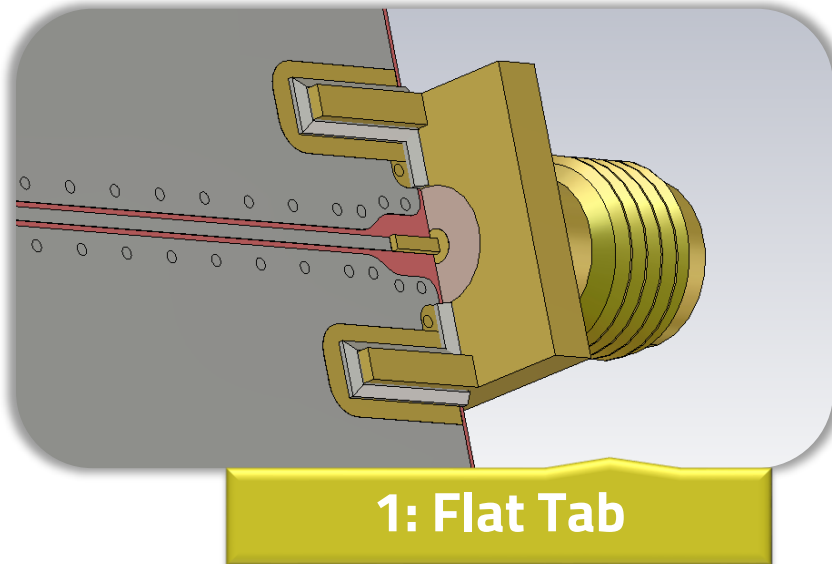
CPWG - ROUND: S-PARAMETER



COMBINATION CONNECTOR & PCB: EXAMPLES

CPWG - ROUND VS FLAT: OVERVIEW

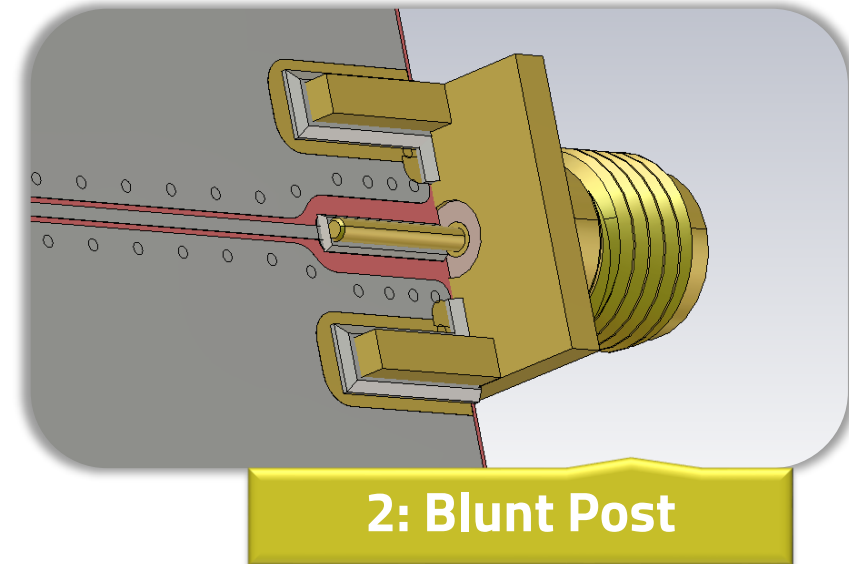
➤ Analysis of 2 different designs



1: Flat Tab

➤ Fits to small traces

- Easy design
- Difficult to solder



2: Blunt Post

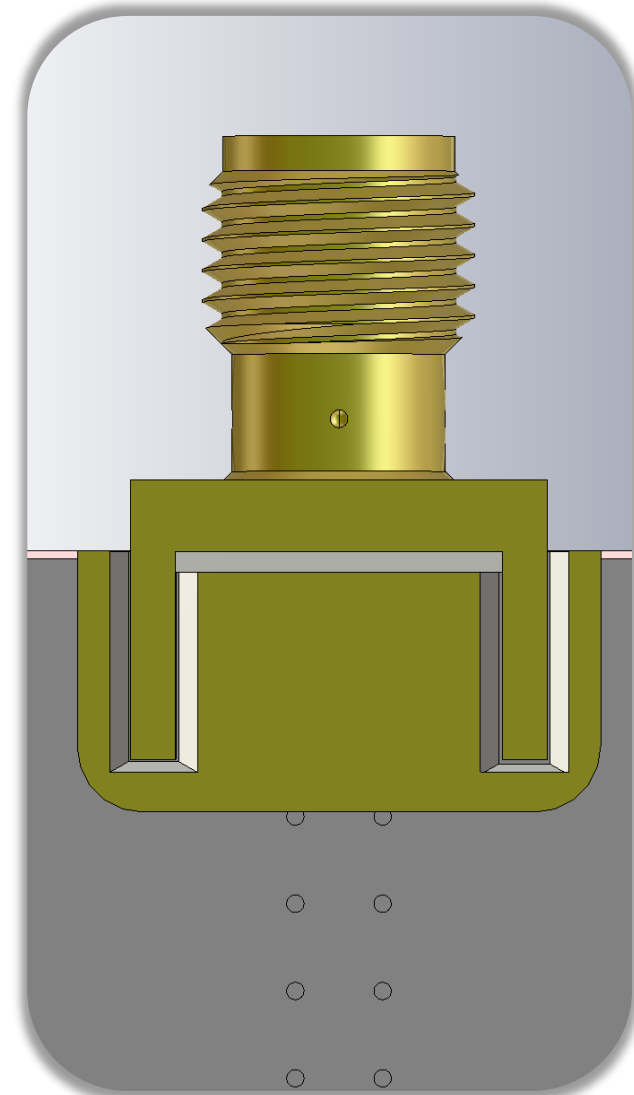
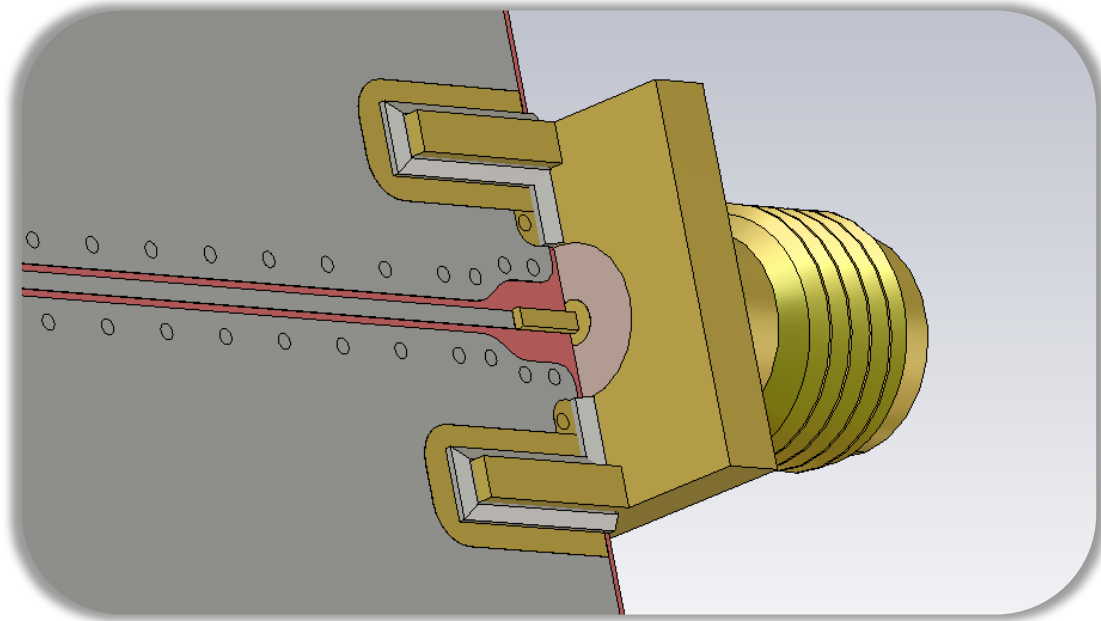
➤ Needs extra solder pad

- Difficult design
- Easy to solder

COMBINATION CONNECTOR & PCB: EXAMPLES

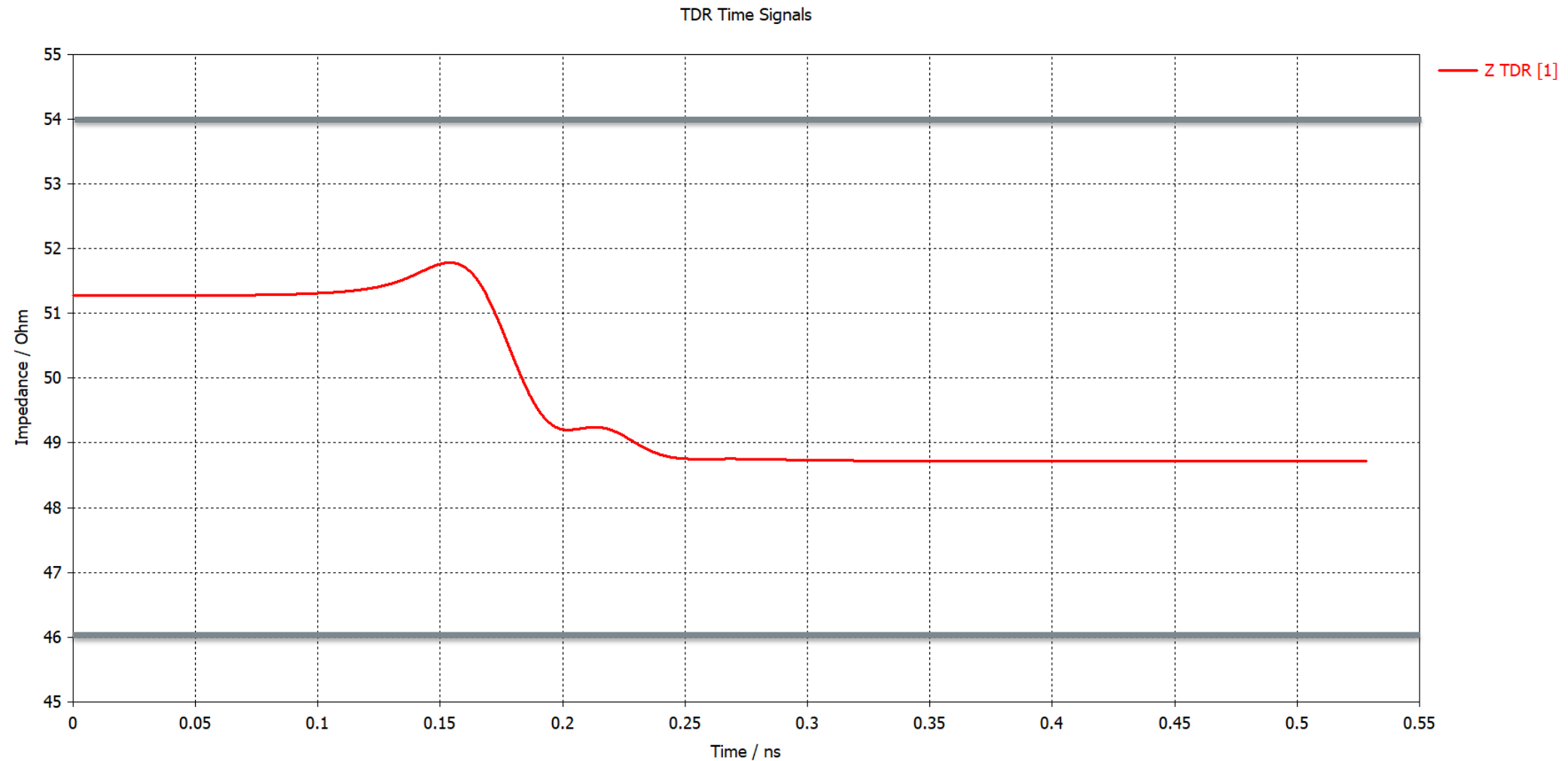
CPWG - FLAT: DESIGN

- Solder pads & several vias
- Very good GND connection
- Optimized RF-Line with taper



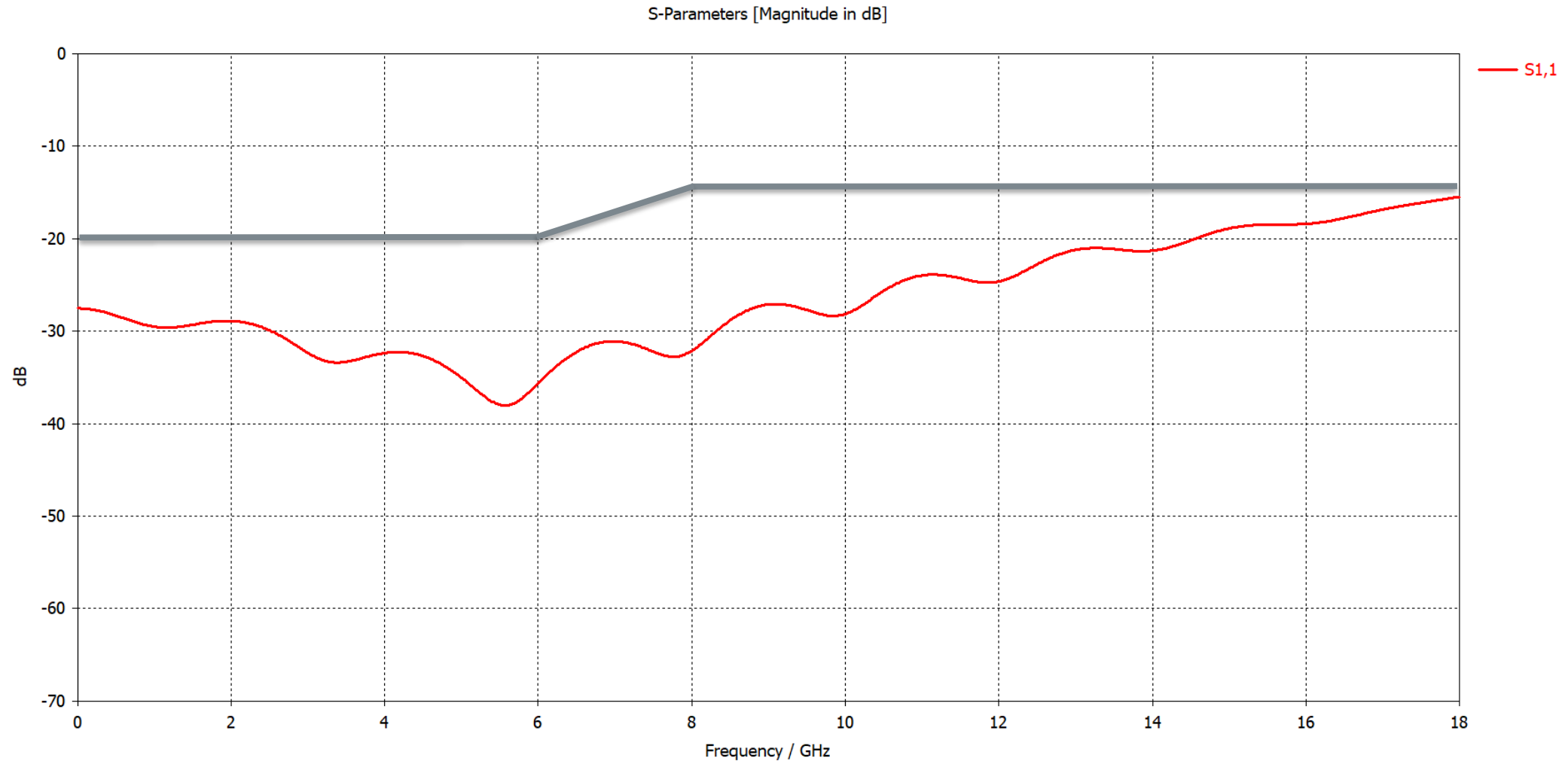
COMBINATION CONNECTOR & PCB: EXAMPLES

CPWG - FLAT: TDR



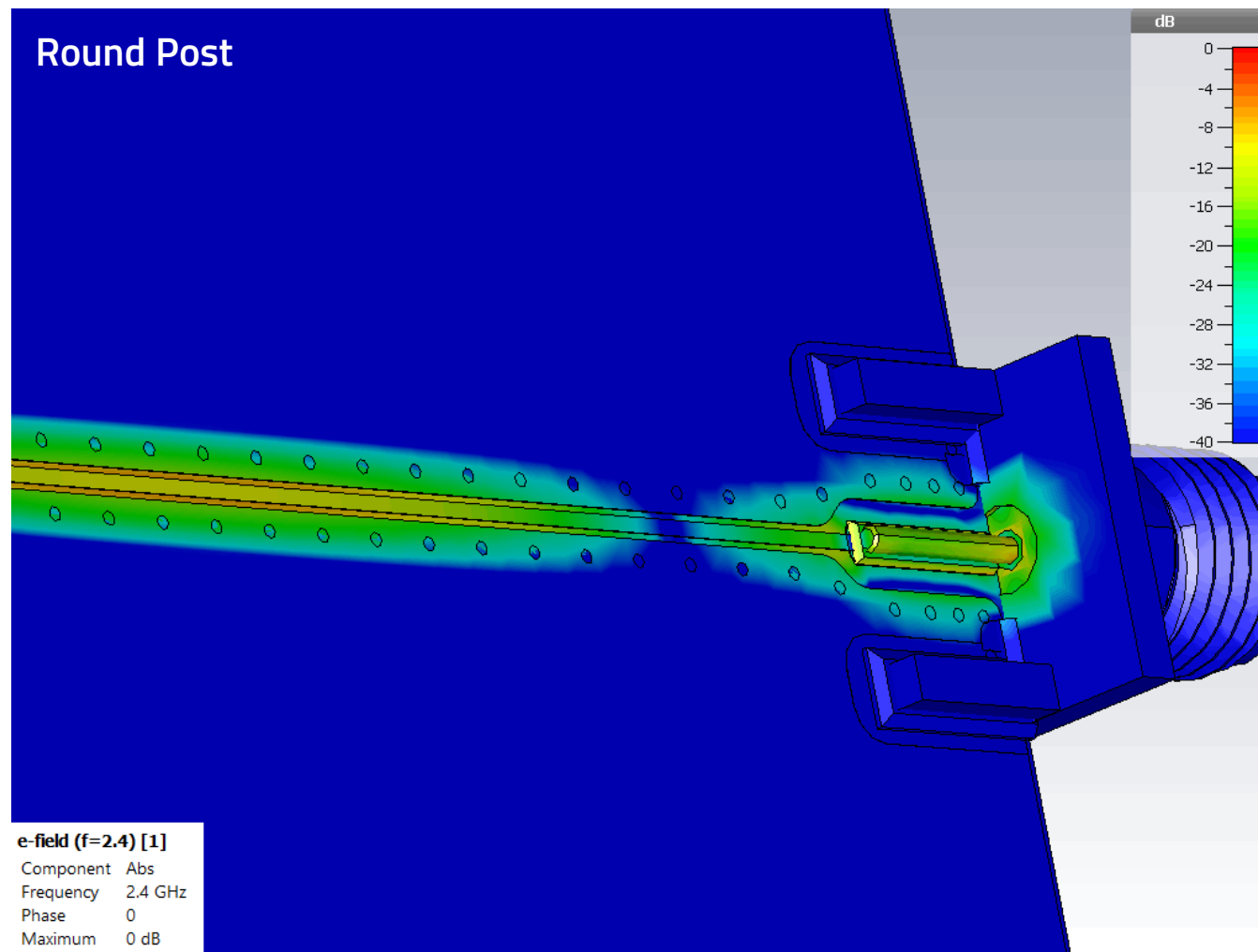
COMBINATION CONNECTOR & PCB: EXAMPLES

CPWG - FLAT: S-PARAMETER



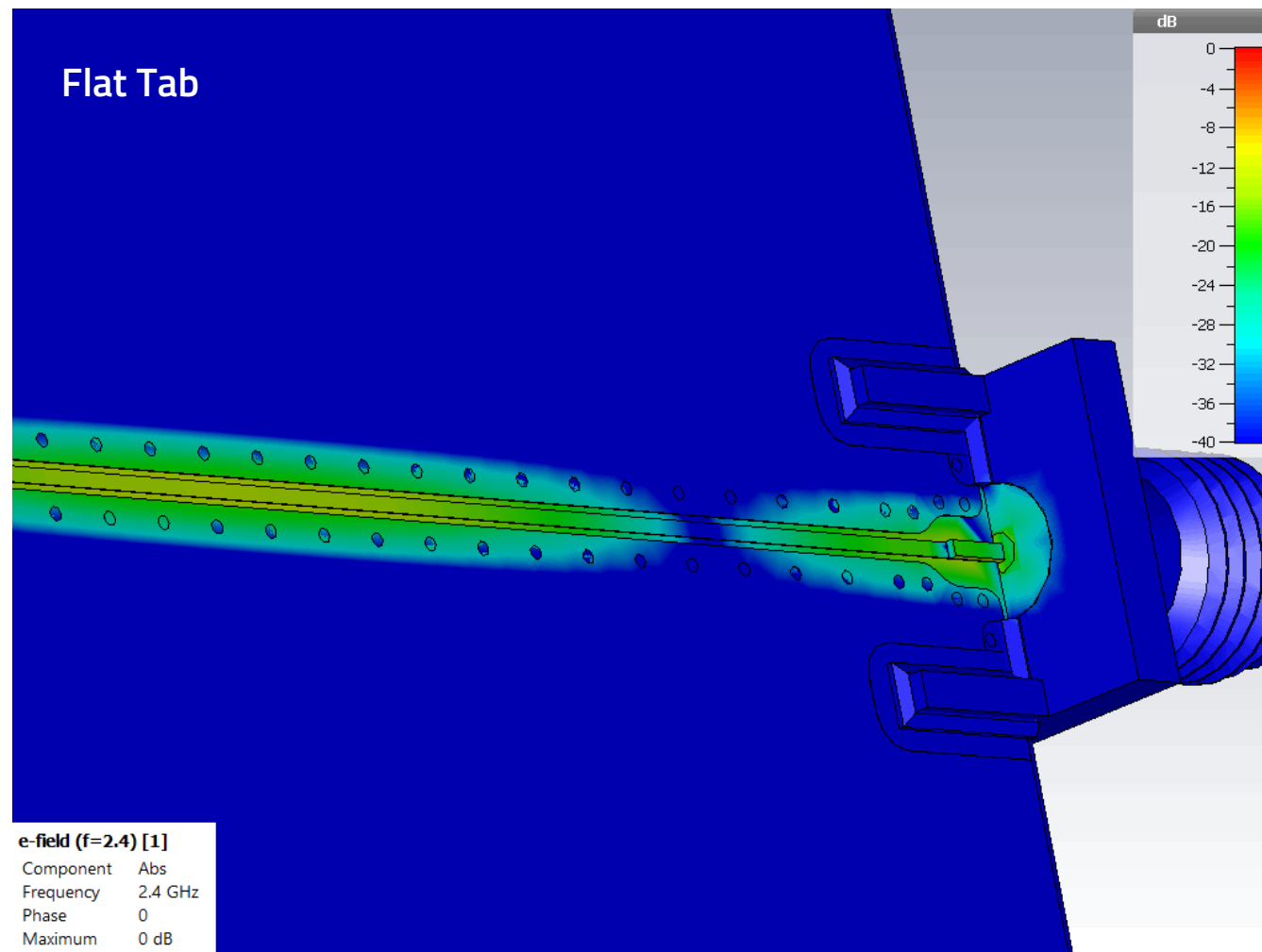
COMBINATION CONNECTOR & PCB: EXAMPLES

CPWG - ROUND: SIMULATION



COMBINATION CONNECTOR & PCB: EXAMPLES

CPWG - FLAT: SIMULATION



COMBINATION CONNECTOR & PCB: EXAMPLES

CPWG - THT: OVERVIEW

➤ Given Task:

- THT **60312102114506**
- 1.55 mm PCB
- 4 Layers – FR4 Core & Prepreg
- CPWG

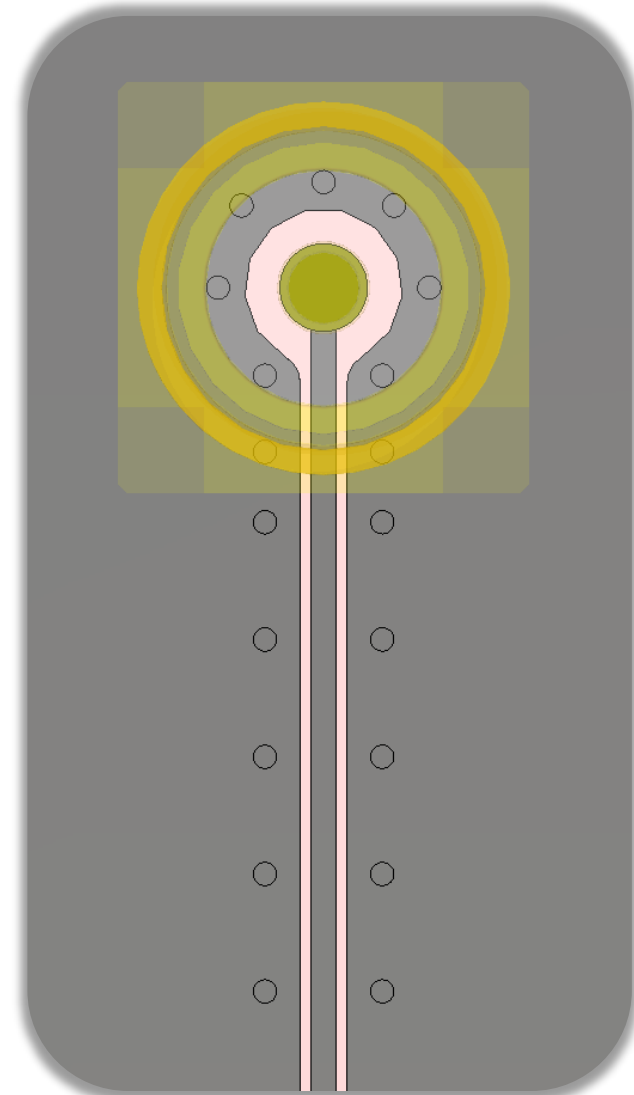
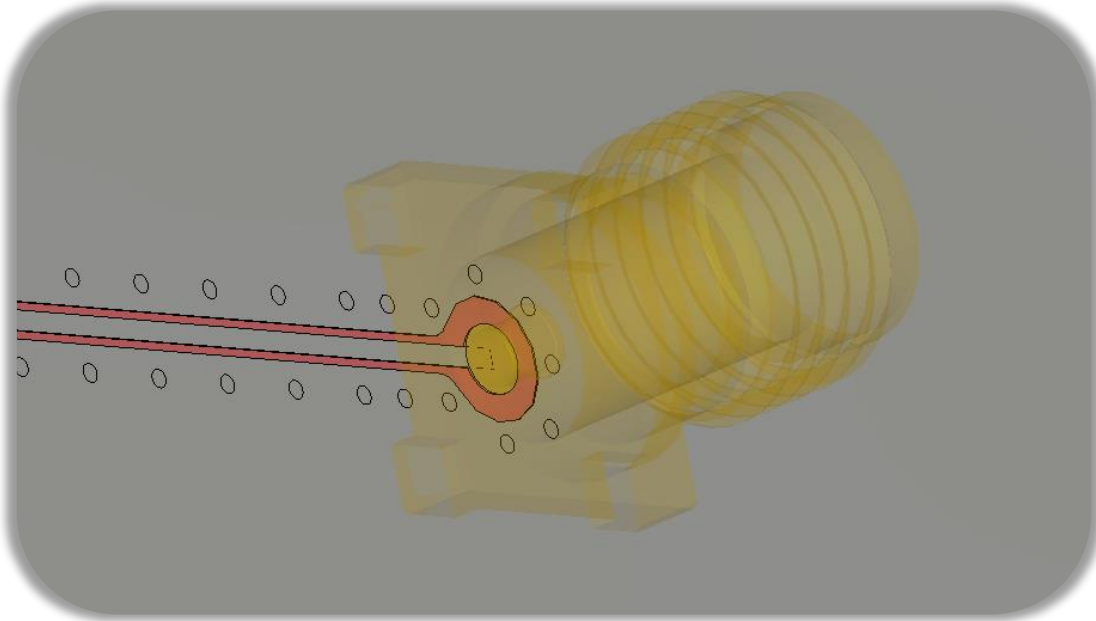
➤ Application: Circuit design



COMBINATION CONNECTOR & PCB: EXAMPLES

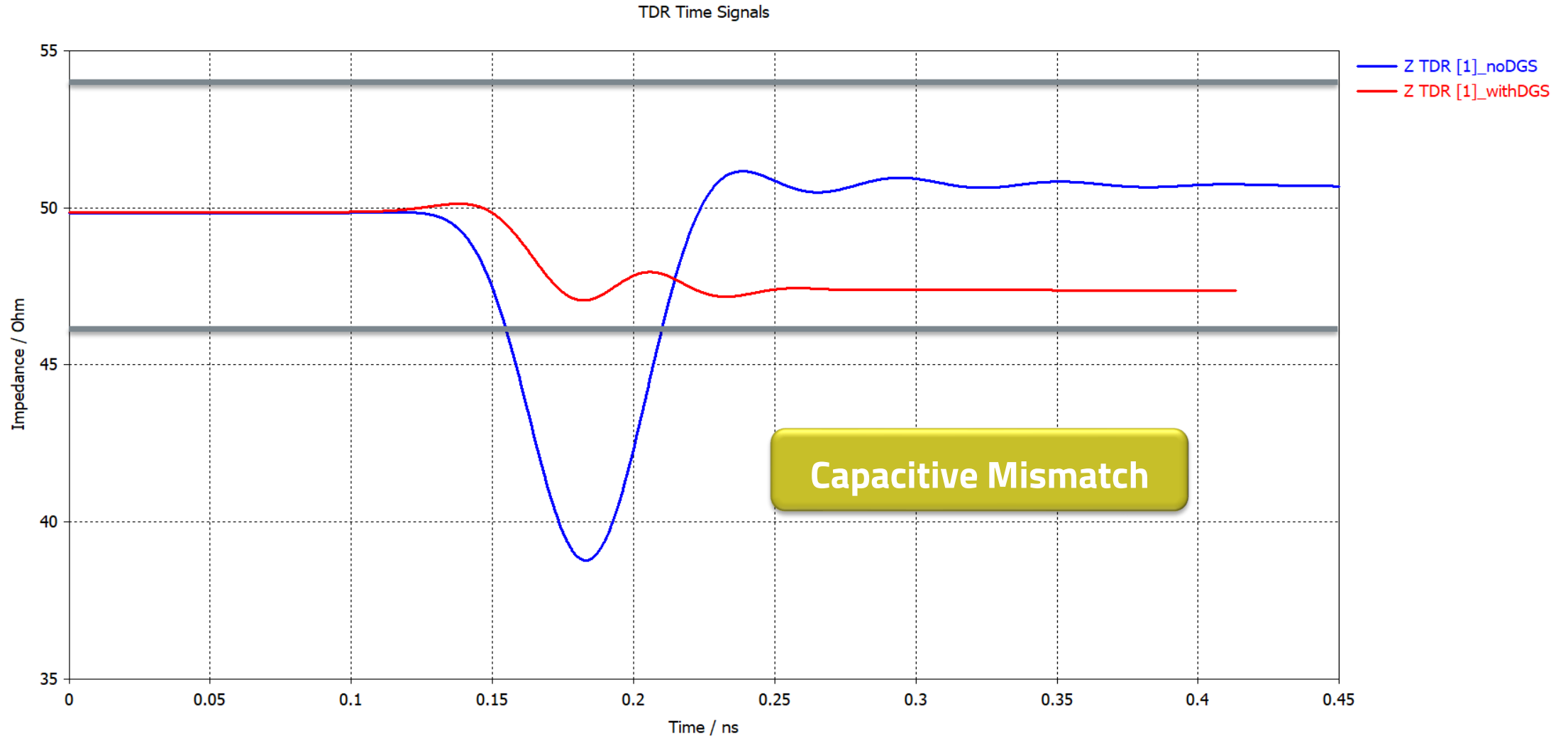
CPWG - THT: DESIGN

- Solderpad & several vias
- Very good GND connection



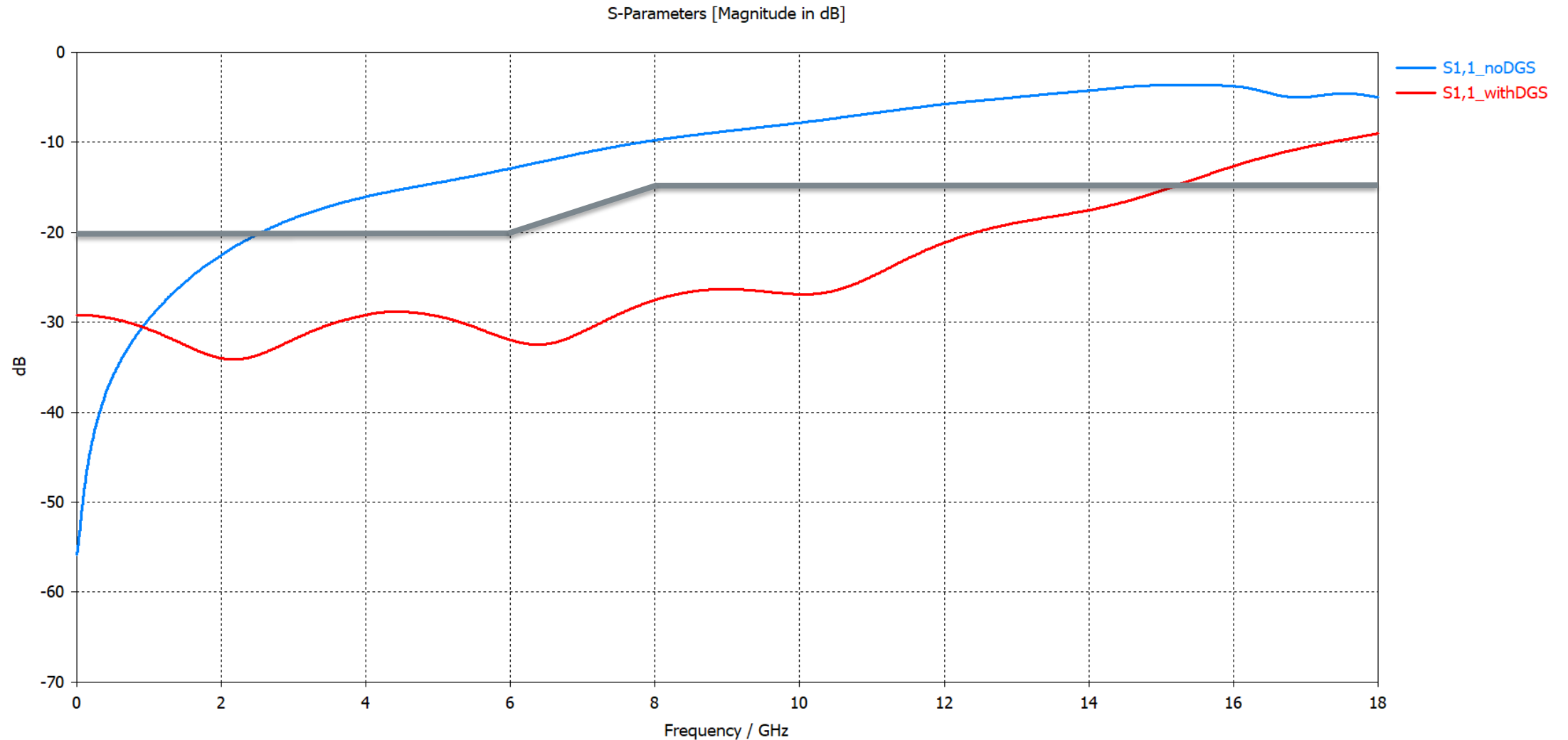
COMBINATION CONNECTOR & PCB: EXAMPLES

CPWG - THT: TDR



COMBINATION CONNECTOR & PCB: EXAMPLES

CPWG - THT: S-PARAMETER

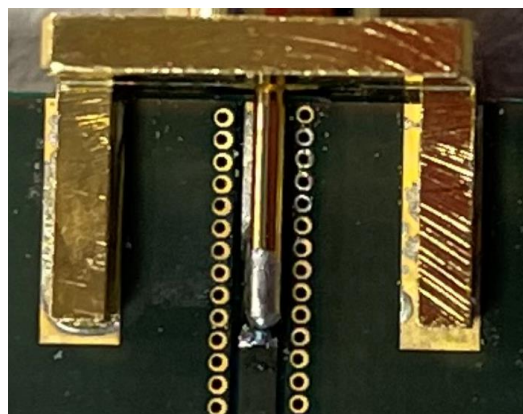


REAL MEASUREMENTS

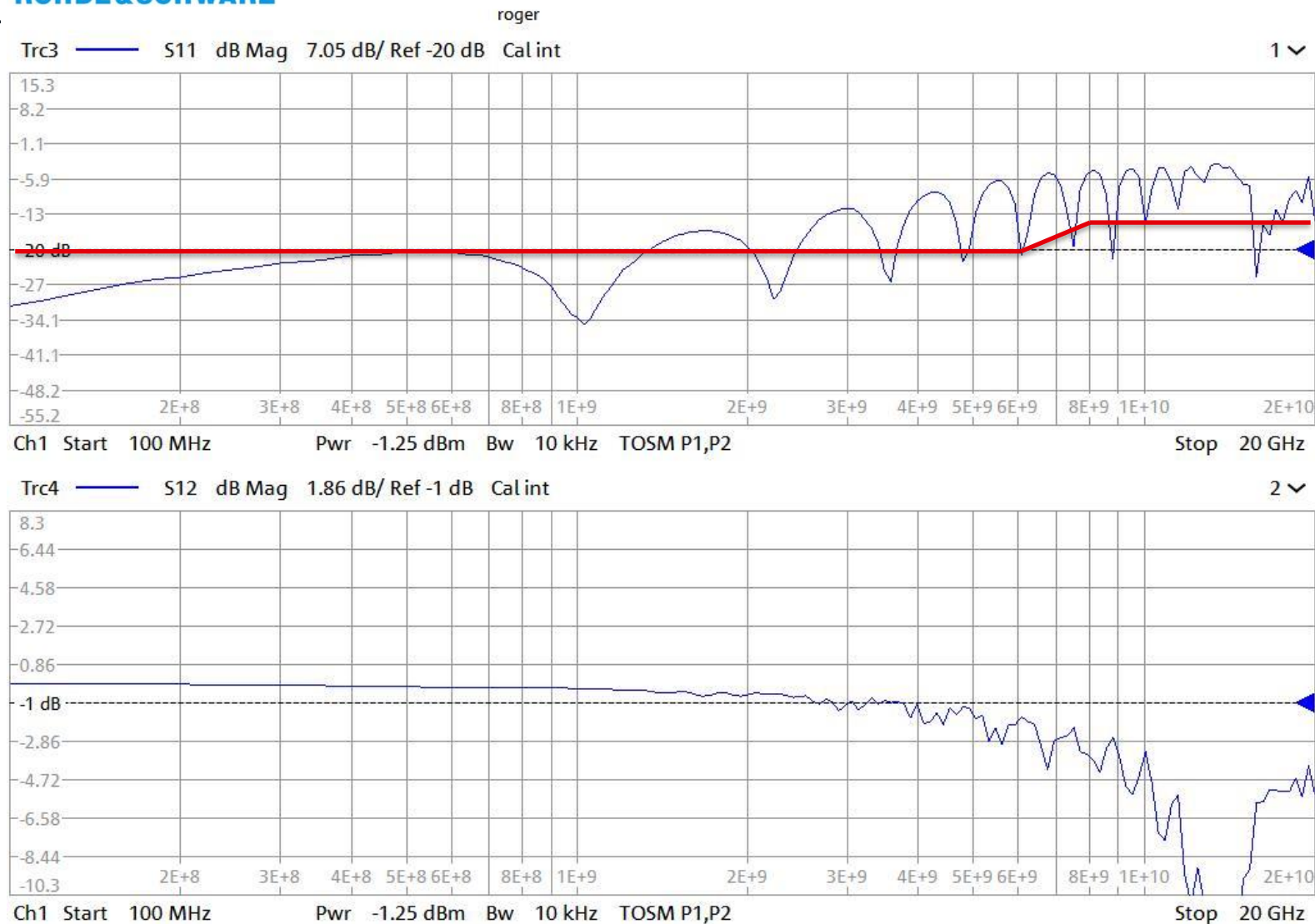
INITIAL WITHOUT TIN



Straight trace
No impedance adjustment



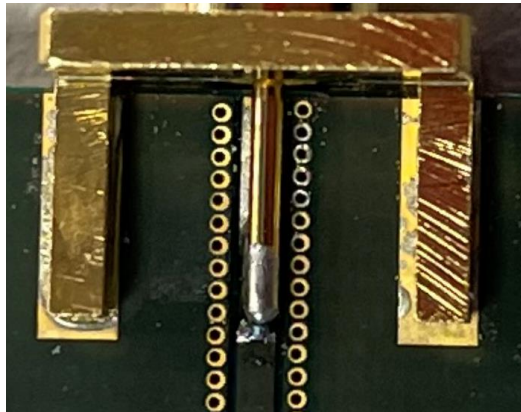
No tin



REAL MEASUREMENTS

INITIAL WITHOUT TIN

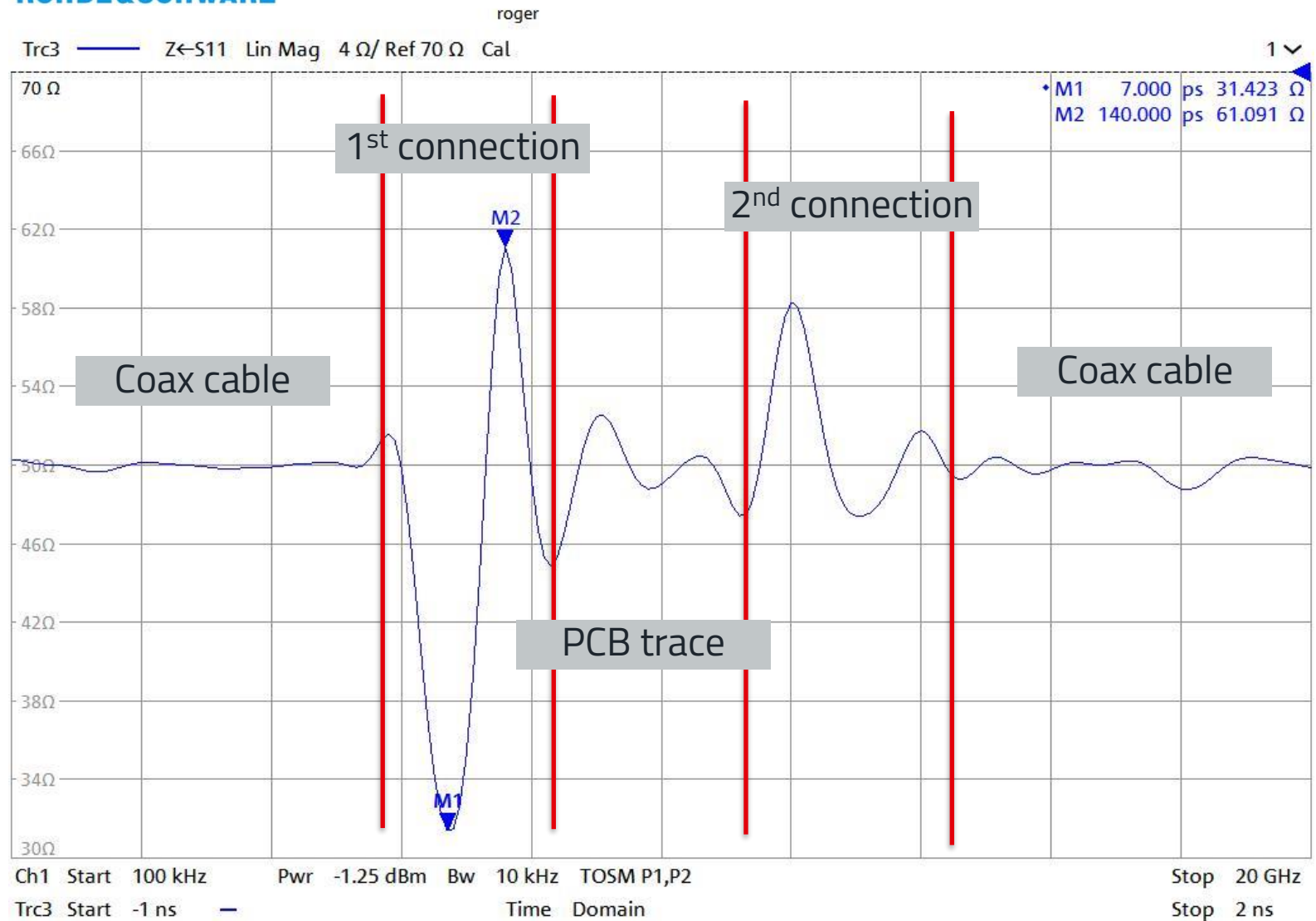
Straight trace
No impedance adjustment



No tin



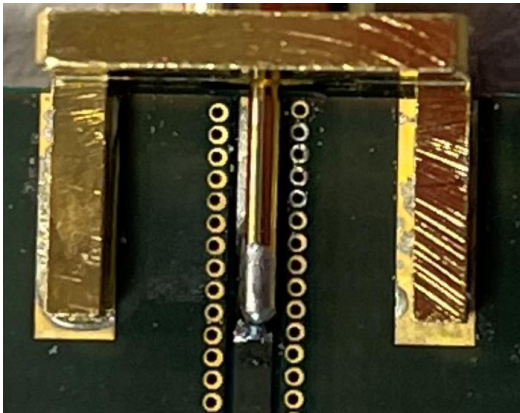
ROHDE & SCHWARZ



REAL MEASUREMENTS

INITIAL WITHOUT TIN

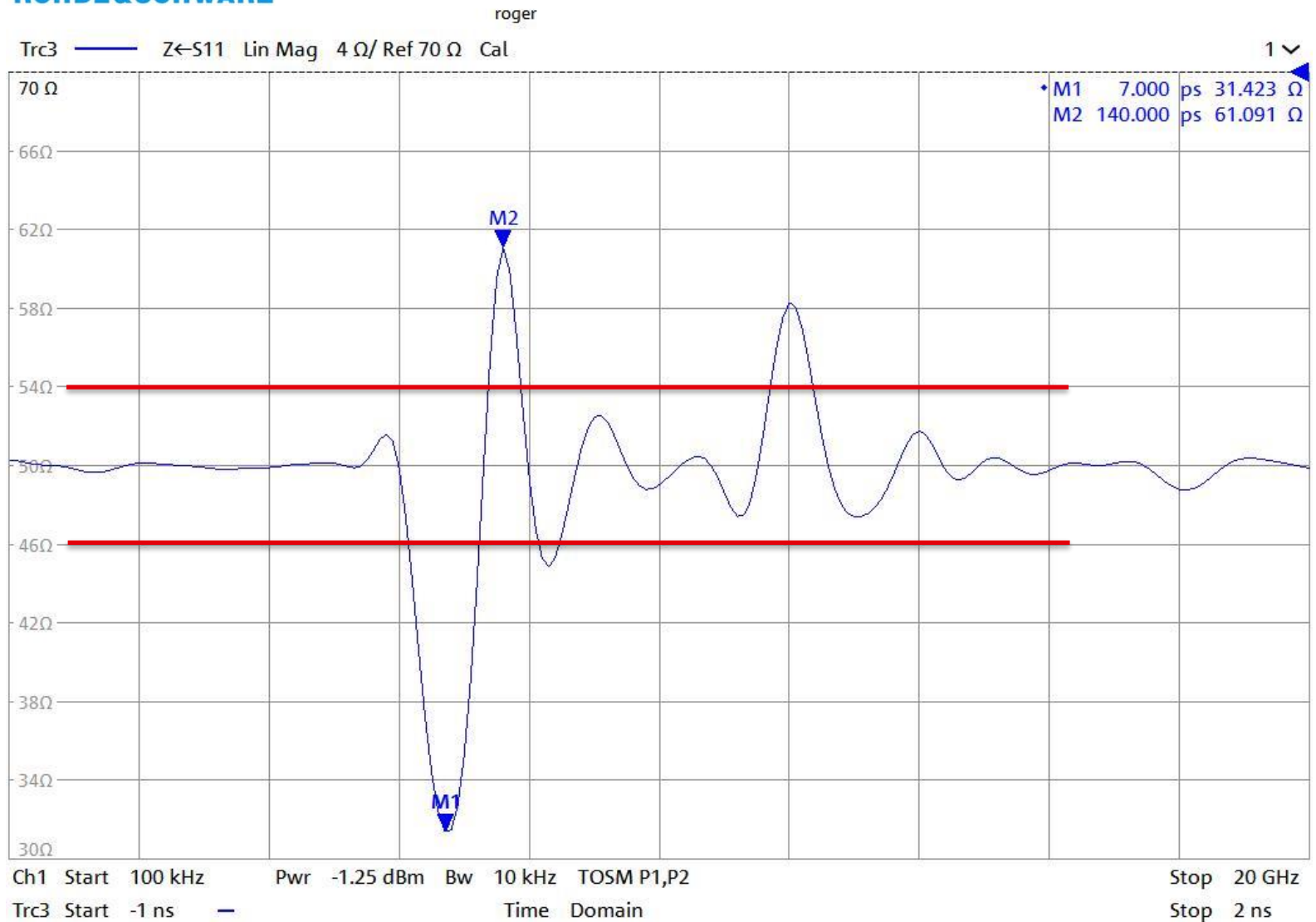
Straight trace
No impedance adjustment



No tin



ROHDE & SCHWARZ

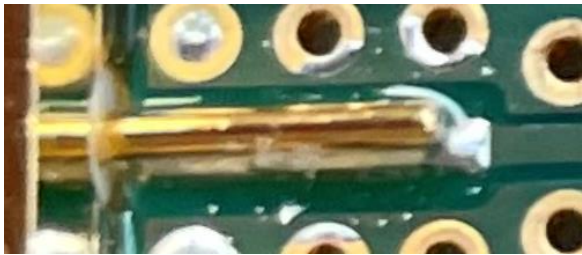


REAL MEASUREMENTS

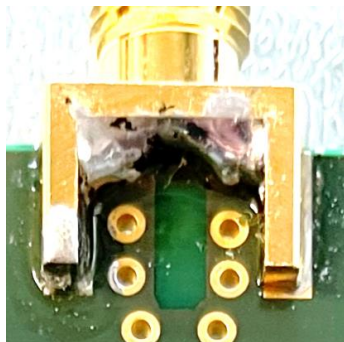
ADJUSTED *SHAPE WITH DGS*

WITH TIN

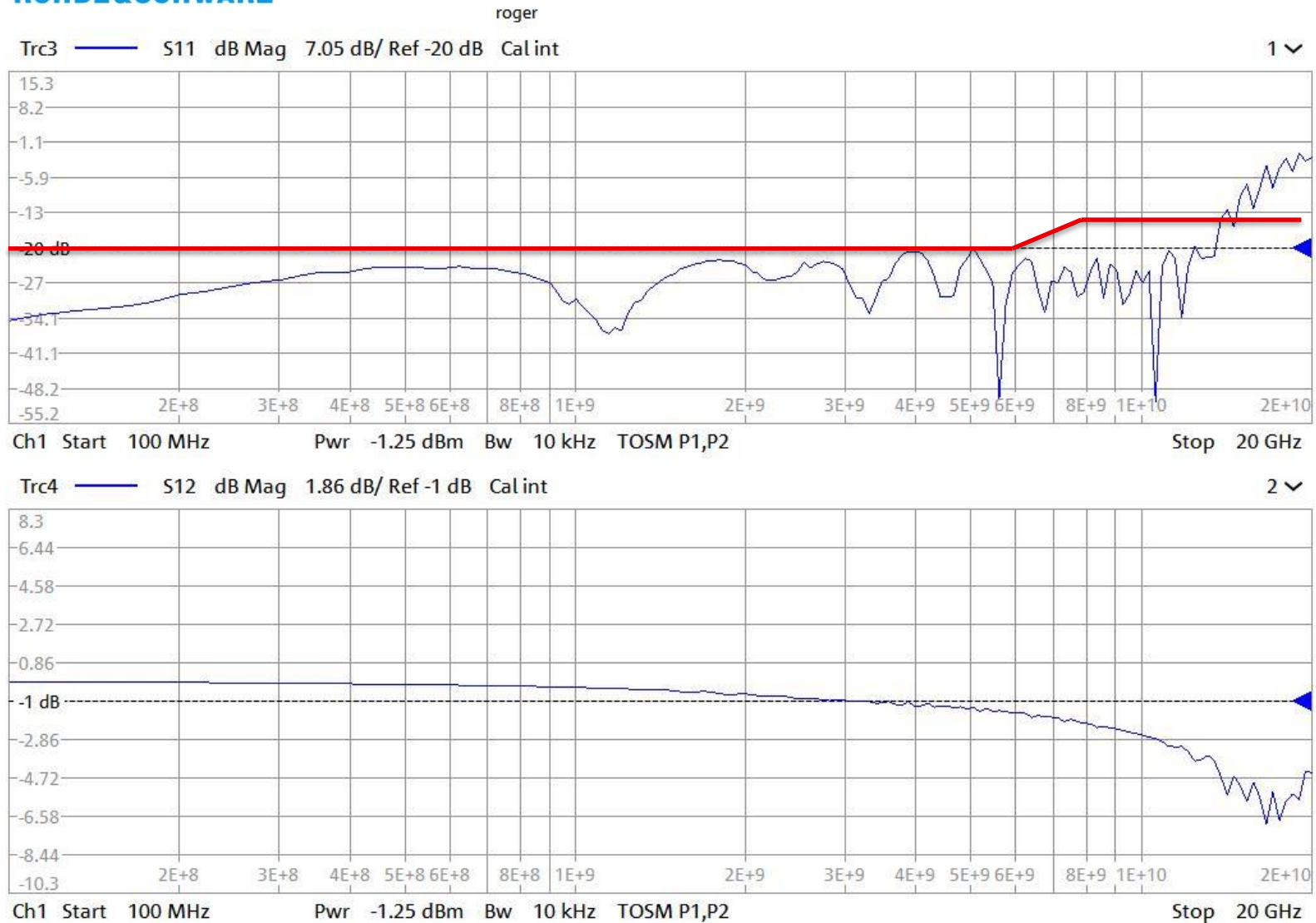
Ground layer adjustment
Adjusted shape



Tin + DGS



ROHDE & SCHWARZ

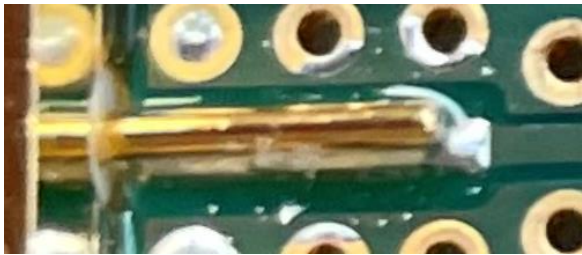


REAL MEASUREMENTS

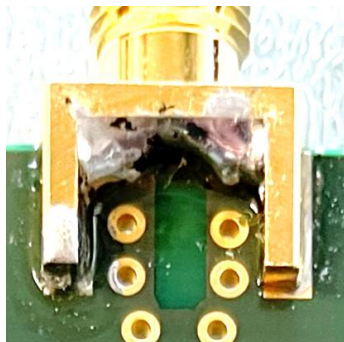
ADJUSTED *SHAPE WITH DGS*

WITH TIN

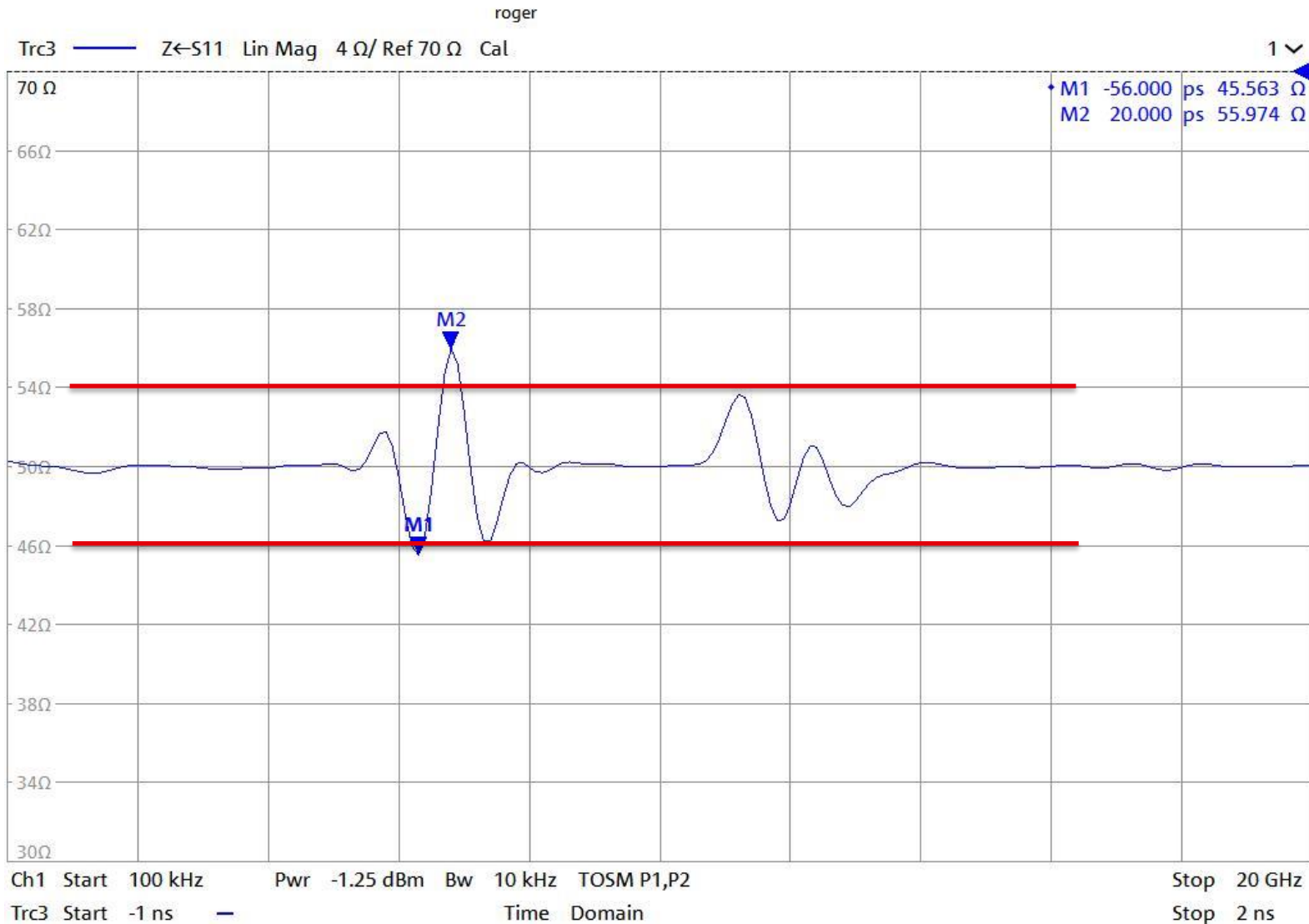
Ground layer adjustment
Adjusted shape



Tin + DGS

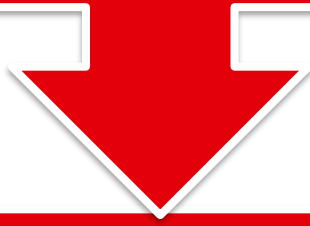


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CONCLUSIONS

Impedance adjustment is important above 500MHz and necessary above 1GHz



It is easy to improve signal integrity:

Use capacitance adaptation, DGS

Avoid soldermask on RF-lines

Use GND via stitching (holes ~ 0,3mm enough)

DZIĘKUJĘ ZA UWAGĘ



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